



# **bq76925 用于 3 至 6 节串联锂离子/锂聚合物电池保护和电量监测的主机控制模拟前端 应用**

## **1 特性**

- 用于主机电芯测量的模拟接口
  - 电芯输入多路开关、电平转换器和缩放器
  - 经过校准的 1.5V/3.0V 低漂移基准电压，可实现精确模数转换
- 用于主机电流测量的模拟接口
  - 可变增益电流感测放大器，能够与 1mΩ 感测电阻搭配运行
- 用于主机温度测量的可切换热敏电阻偏置输出
- 阈值可动态调节的过流比较器
  - 向主机报告潜在的过流故障
  - 在连接负载后唤醒主机
- 集成电芯均衡场效应晶体管 (FET)
  - 单独主机控制
  - 每个电芯的均衡电流为 50mA
- 支持电芯感测线路开路检测
- 集成有用于为微控制器或 LED 供电的 3.3V 稳压器
- 用于主机通信的 I<sup>2</sup>C 接口
  - 可选数据包循环冗余校验 (CRC)，用于确保通信的稳健性
- 电源电压范围：4.2V 至 26.4V
- 低功耗
  - 正常模式下的电流典型值为 40μA
  - 休眠模式下的电流最大值为 1.5μA
- 采用 20 引脚散热薄型小外形尺寸 (HTSSOP) 或 24 引脚超薄四方扁平无引线 (VQFN) 封装

## **2 应用**

- 锂离子电池组主保护
  - 无绳电动工具
  - 轻型电动车辆（电动自行车和电动踏板车等）
  - 不间断电源 (UPS) 系统
  - 医疗设备
  - 便携式测试设备

## **3 说明**

Bq76925 是一款由主机控制的模拟前端 (AFE)，它是一套完备电池组监视、均衡和保护系统的组成部分，这套系统适用于 3、4、5 或 6 节串联锂离子和锂聚合物电池。bq76925 器件通过一个主机控制器轻松监视各电芯电压、电池组电流及温度。主机可利用这些信息确定不安全或故障运行情况，例如过压、欠压、过热、过流、电芯电量失衡、充电状态和健康状况。

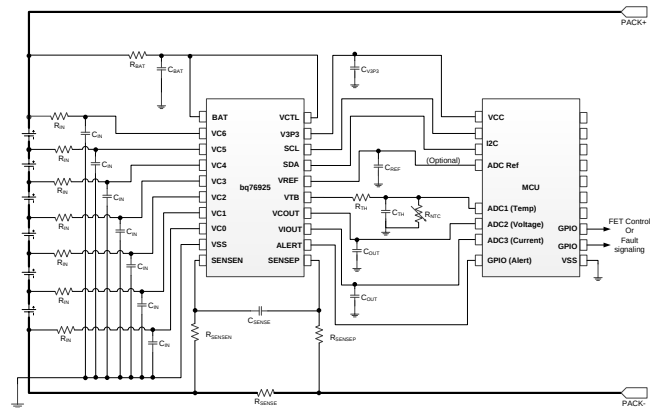
电芯输入电压经电平转换、多路选择、缩放后输出，以供主机 ADC 测量。一个专用引脚提供了经校准的低漂移基准电压，从而实现精确测量。

**器件信息<sup>(1)</sup>**

| 器件型号    | 封装         | 封装尺寸（标称值）       |
|---------|------------|-----------------|
| bq76925 | TSSOP (20) | 4.00mm x 4.00mm |
| bq76925 | VQFN (24)  | 6.50mm x 4.40mm |

(1) 要了解所有可用封装，请见数据表末尾的可订购产品附录。

**简化电路原理图**



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## 4 修订历史记录

注：之前版本的页码可能与当前版本有所不同。

| Changes from Revision C (July 2015) to Revision D                                                     | Page |
|-------------------------------------------------------------------------------------------------------|------|
| Added test condition $n = 1 - 5$ at $25^{\circ}\text{C}$ and MAX value for $I_{\text{VCN}}$ parameter | 6    |
| 已添加 接收文档更新通知部分                                                                                        | 33   |

| Changes from Revision B (December 2011) to Revision C                                                                                                                                                                                                                             | Page |
|-----------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|------|
| 已添加 ESD 额定值表, 特性 说明 部分, 器件功能模式, 应用和实施部分, 电源相关建议部分, 布局部分, 器件和文档支持部分以及机械、封装和可订购信息部分                                                                                                                                                                                                 | 1    |
| 已将内容移至新的部分, 并已添加相应部分、图、表和文档的超链接。                                                                                                                                                                                                                                                  | 1    |
| Moved $R_{\text{BAT}}$ , $C_{\text{BAT}}$ , $R_{\text{IN}}$ , $C_{\text{IN}}$ , $R_{\text{SENSE}}$ , $R_{\text{SENSEP}}$ , $C_{\text{SENSE}}$ , $R_{\text{VCTL}}$ , $C_{\text{V3P3}}$ , $C_{\text{REF}}$ , and $C_{\text{OUT}}$ table rows to <a href="#">Design Requirements</a> | 5    |

| Changes from Revision A (July 2011) to Revision B | Page |
|---------------------------------------------------|------|
| Added 24-pin QFN (RGE) Package to Production Data | 3    |

| Changes from Original (July 2011) to Revision A           | Page |
|-----------------------------------------------------------|------|
| Changed literature number to Rev A for ProductMix release | 4    |

## 5 说明（续）

外部感测电阻两端的电压经过放大后输出到主机 **ADC**，以进行充放电电流测量。两种增益设置使得器件能够搭配多种感测电阻值使用，以便在各种电池组电流下运行。

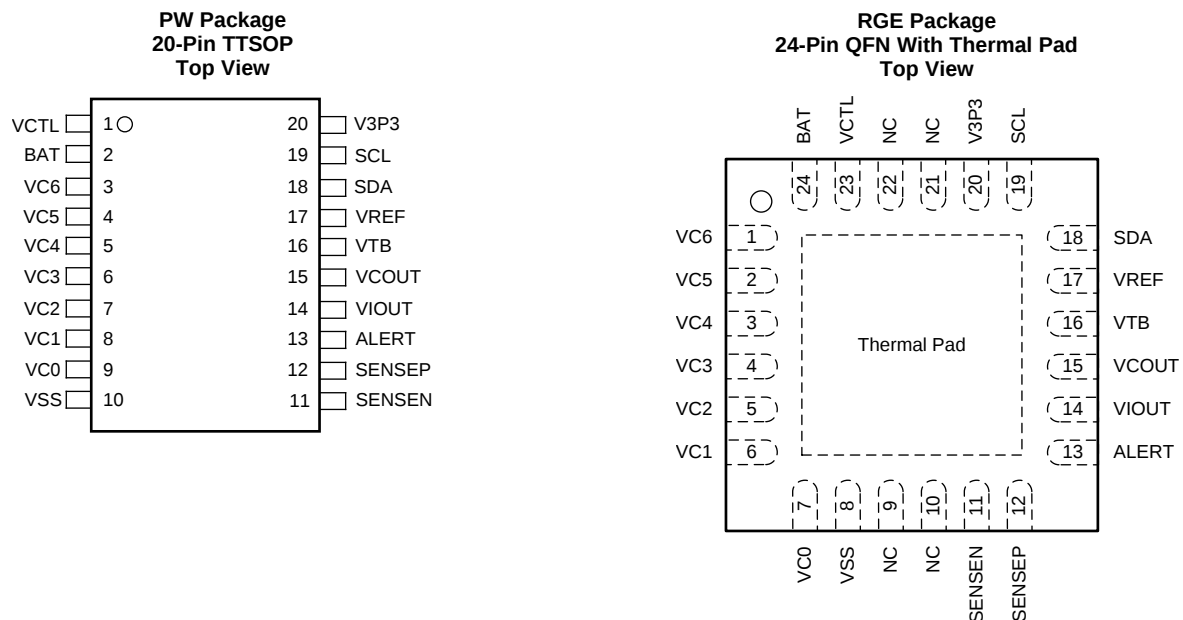
为了使主机能够进行温度测量，**AFE** 提供了一个单独的输出引脚，用于为外部热敏电阻网络提供偏置。该输出可在主机的控制下打开/关闭，从而将功耗降至最小。

**bq76925** 器件包含一个阈值可动态选择的比较器，用于监视电流。比较器结果通过开漏输出驱动，以在超过阈值时向主机报警。该特性可用于在连接负载后唤醒主机，或者向主机报告潜在的故障情况。

**bq76925** 器件集成了完全受主机控制的电芯均衡 **FET**。均衡电流通过外部电阻设置，最高可设为 **50mA**。可以将这些相同的 **FET** 与电芯电压测量结果搭配使用，以检测电芯检测线路上的开路情况。

主机通过 **I<sup>2</sup>C** 接口与 **AFE** 通信。可以选择使用数据包 **CRC** 以确保通信的稳健性。可以通过 **I<sup>2</sup>C** 接口将器件置于低电流休眠模式，并且可以通过上拉 **ALERT** 引脚唤醒器件。

## 6 Pin Configuration and Functions



### Pin Functions

| NAME   | PIN NO. |               | TYPE           | DESCRIPTION                                    |
|--------|---------|---------------|----------------|------------------------------------------------|
|        | TSSOP   | VQFN          |                |                                                |
| VCTL   | 1       | 23            | Output         | 3.3-V Regulator control voltage <sup>(1)</sup> |
| ALERT  | 13      | 13            | Output         | Overcurrent alert (open drain)                 |
| BAT    | 2       | 24            | Power          | Supply voltage, tied to most positive cell     |
| NC     | —       | 9, 10, 21, 22 | —              | No Connection (leave open)                     |
| SCL    | 19      | 19            | Input          | I <sup>2</sup> C Clock (open drain)            |
| SDA    | 18      | 18            | Input / Output | I <sup>2</sup> C Data (open drain)             |
| SENSEN | 11      | 11            | Input          | Negative current sense                         |
| SENSEP | 12      | 12            | Input          | Positive current sense                         |
| V3P3   | 20      | 20            | Output         | 3.3-V Regulator                                |
| VC6    | 3       | 1             | Input          | Sense voltage for most positive cell           |

(1) When a bypass FET is used to supply the regulated 3.3-V load current, VCTL automatically adjusts to keep V3P3 = 3.3 V. If VCTL is tied to BAT, the load current is supplied through V3P3.

### Pin Functions (continued)

| NAME  | PIN NO. |      | TYPE   | DESCRIPTION                                  |
|-------|---------|------|--------|----------------------------------------------|
|       | TSSOP   | VQFN |        |                                              |
| VC5   | 4       | 2    | Input  | Sense voltage for second most positive cell  |
| VC4   | 5       | 3    | Input  | Sense voltage for third most positive cell   |
| VC3   | 6       | 4    | Input  | Sense voltage for fourth most positive cell  |
| VC2   | 7       | 5    | Input  | Sense voltage for fifth most positive cell   |
| VC1   | 8       | 6    | Input  | Sense voltage for least positive cell        |
| VC0   | 9       | 7    | Input  | Sense voltage for negative end of cell stack |
| VCOUT | 15      | 15   | Output | Cell measurement voltage                     |
| VIOUT | 14      | 14   | Output | Current measurement voltage                  |
| VREF  | 17      | 17   | Output | Reference voltage for ADC                    |
| VSS   | 10      | 8    | Power  | Ground                                       |
| VTB   | 16      | 16   | Output | Bias voltage for thermistor network          |

## 7 Specifications

### 7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted) <sup>(1)</sup>

|                  |                        |                                                                            | MIN  | MAX     | UNIT |
|------------------|------------------------|----------------------------------------------------------------------------|------|---------|------|
| V <sub>BAT</sub> | Supply voltage         | BAT                                                                        | −0.3 | 36      | V    |
| V <sub>I</sub>   | Input voltage          | Cell input differential, VC <sub>n</sub> to VC <sub>n+1</sub> , n = 0 to 5 | −0.3 | 9       | V    |
|                  |                        | Cell input, VC <sub>n</sub> , n = 1 to 6                                   | −0.3 | (6 × n) |      |
|                  |                        | BAT to VC6 differential                                                    | −10  | 10      |      |
|                  |                        | VC0 <sup>(2)</sup>                                                         | −3   | 3       |      |
|                  |                        | SENSEP, SENSEN                                                             | −3   | 3       |      |
|                  |                        | SCL, SDA                                                                   | −0.3 | 6       |      |
| V <sub>O</sub>   | Output voltage         | VCOUT, VIOUT, VREF                                                         | −0.3 | 3.6     | V    |
|                  |                        | VTB, V3P3                                                                  | −0.3 | 7       |      |
|                  |                        | ALERT                                                                      | −0.3 | 30      |      |
|                  |                        | VCTL                                                                       | −0.3 | 36      |      |
| I <sub>CB</sub>  | Cell balancing current |                                                                            |      | 70      | mA   |
| I <sub>IN</sub>  | Cell input current     |                                                                            | −25  | 70      | mA   |
| T <sub>stg</sub> | Storage temperature    |                                                                            | −65  | 150     | °C   |

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) Negative voltage swings on VC0 in the absolute maximum range can cause unwanted circuit behavior and should be avoided.

### 7.2 ESD Ratings

|                    |                         |                                                                                | VALUE | UNIT |
|--------------------|-------------------------|--------------------------------------------------------------------------------|-------|------|
| V <sub>(ESD)</sub> | Electrostatic discharge | Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>              | ±2000 | V    |
|                    |                         | Charged-device model (CDM), per JEDEC specification JESD22-C101 <sup>(2)</sup> | ±500  |      |

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

### 7.3 Recommended Operating Conditions<sup>(1)</sup>

|                   |                                 |                | MIN                                               | NOM                        | MAX     | UNIT |            |
|-------------------|---------------------------------|----------------|---------------------------------------------------|----------------------------|---------|------|------------|
| Supply voltage    |                                 | BAT            |                                                   | 4.2                        | 26.4    | V    |            |
| V <sub>I</sub>    | Input voltage                   |                | Cell input differential, VCn to VCn+1, n = 0 to 5 |                            | 1.4     | 4.4  | V          |
|                   |                                 |                | Cell input, VCn, n = 1 to 6                       |                            | 4.4 × n |      | V          |
|                   |                                 |                | BAT to VC6 differential                           |                            | −8      | 8    | V          |
|                   |                                 |                | VC0, SENSEN                                       |                            | 0       |      | V          |
|                   |                                 |                | SENSEP                                            |                            | −125    | 375  | mV         |
|                   |                                 |                | SCL, SDA                                          |                            | 0       | 5.5  | V          |
|                   |                                 |                | V3P3                                              | Backfeeding <sup>(2)</sup> | 5.5     |      | V          |
|                   |                                 |                | ALERT                                             | Wakeup function            | 0       | 26.4 | V          |
|                   | V <sub>O</sub>                  | Output voltage |                                                   | VCOUT, VIOUT               |         | 0    | V3P3 + 0.2 |
| VREF              |                                 |                |                                                   | REFSEL = 0                 |         | 1.5  | V          |
|                   |                                 |                |                                                   | REFSEL = 1                 |         | 3    | V          |
| VTB               |                                 |                |                                                   | 5.5                        |         | V    |            |
| V3P3              |                                 |                |                                                   | Regulating                 | 3.3     |      | V          |
| VCTL              |                                 |                |                                                   | 0.8                        | 26.4    | V    |            |
| ALERT             |                                 | Alert function | 0                                                 | 5.5                        | V       |      |            |
| I <sub>CB</sub>   | Cell balancing current          |                | 0                                                 | 50                         | mA      |      |            |
| T <sub>A</sub>    | Operating free-air temperature  |                | −25                                               | 85                         | °C      |      |            |
| T <sub>FUNC</sub> | Functional free-air temperature |                | −40                                               | 100                        | °C      |      |            |

(1) All voltages are relative to VSS, except “Cell input differential.”

(2) Internal 3.3-V regulator may be overridden (that is, backfed) by applying an external voltage larger than the regulator voltage.

### 7.4 Thermal Information

| THERMAL METRIC <sup>(1)</sup> |                                              | bq76925    |            | UNIT |
|-------------------------------|----------------------------------------------|------------|------------|------|
|                               |                                              | PW (TSSOP) | RGE (VQFN) |      |
|                               |                                              | 20 PINS    | 24 PINS    |      |
| R <sub>θJA</sub>              | Junction-to-ambient thermal resistance       | 97.5       | 36         | °C/W |
| R <sub>θJC (top)</sub>        | Junction-to-case (top) thermal resistance    | 31.7       | 38.6       | °C/W |
| R <sub>θJB</sub>              | Junction-to-board thermal resistance         | 48.4       | 14         | °C/W |
| ψ <sub>JT</sub>               | Junction-to-top characterization parameter   | 1.5        | 0.6        | °C/W |
| ψ <sub>JB</sub>               | Junction-to-board characterization parameter | 47.9       | 14         | °C/W |
| R <sub>θJC (bot)</sub>        | Junction-to-case (bottom) thermal resistance | n/a        | 4.6        | °C/W |

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

## 7.5 Electrical Characteristics: Supply Current

BAT = 4.2 to 26.4 V, VCn = 1.4 to 4.4, T<sub>A</sub> = –25°C to +85°C

Typical values stated where T<sub>A</sub> = 25°C and BAT = 21.6 V (unless otherwise noted)

| PARAMETER         |                                       | TEST CONDITION                                                                                                       |                   | MIN | TYP | MAX   | UNIT |
|-------------------|---------------------------------------|----------------------------------------------------------------------------------------------------------------------|-------------------|-----|-----|-------|------|
| I <sub>DD1</sub>  | Normal mode supply current            | All device functions enabled<br>All pins unloaded<br>SDA and SCL high                                                |                   |     | 40  | 48    | μA   |
| I <sub>DD2</sub>  | Standby mode 1 supply current         | V3P3 and overcurrent monitor enabled<br>All pins unloaded<br>All other device functions disabled<br>SDA and SCL high |                   |     | 14  | 17    | μA   |
| I <sub>DD3</sub>  | Standby mode 2 supply current         | V3P3 enabled<br>All pins unloaded<br>All device functions disabled<br>SDA and SCL high                               |                   |     | 12  | 14    | V    |
| I <sub>DD4</sub>  | Sleep mode supply current             | V3P3 disabled<br>All pins unloaded<br>All device functions disabled<br>SDA and SCL low                               |                   |     | 1   | 1.5   | μA   |
| I <sub>VCn</sub>  | Input current for selected cell       | All cell voltages equal                                                                                              | n = 6             |     | 2.4 | 2.7   | μA   |
|                   |                                       | Cell balancing disabled                                                                                              | n = 1 – 5         |     |     | < 0.5 |      |
|                   |                                       | Open cell detection disabled during cell voltage monitoring                                                          | n = 1 – 5 at 25°C |     |     | < 0.3 |      |
| ΔI <sub>VCn</sub> | Cell to cell input current difference | All cell voltages equal<br>Cell balancing disabled<br>Open cell detection disabled                                   |                   |     |     | < 0.2 | μA   |

## 7.6 Internal Power Control (Startup and Shutdown)

| PARAMETER             |                                                  | TEST CONDITION                   |                                              | MIN | TYP | MAX | UNIT |
|-----------------------|--------------------------------------------------|----------------------------------|----------------------------------------------|-----|-----|-----|------|
| V <sub>POR</sub>      | Power on reset voltage                           | Measured at BAT pin              | Initial BAT < 1.4 VBAT rising <sup>(1)</sup> | 4.3 | 4.5 | 4.7 | V    |
|                       |                                                  |                                  | Initial BAT > 1.4 VBAT rising <sup>(1)</sup> | 6.5 | 7   | 7.5 | V    |
| V <sub>SHUT</sub>     | Shutdown voltage <sup>(2)</sup>                  | Measured at BAT pin, BAT falling |                                              |     |     | 3.6 | V    |
| t <sub>POR</sub>      | Time delay after POR before I2C comms allowed    | CV3P3 = 4.7 μF                   |                                              |     |     | 1   | ms   |
| V <sub>WAKE</sub>     | Wakeup voltage                                   | Measured at ALERT pin            |                                              | 0.8 |     | 2   | V    |
| t <sub>WAKE_PLS</sub> | Wakeup signal pulse width                        |                                  |                                              | 1   |     | 5   | μs   |
| t <sub>WAKE_DLY</sub> | Time delay after wakeup before I2C comms allowed | CV3P3 = 4.7 μF                   |                                              |     |     | 1   | ms   |

(1) Initial power up will start with BAT < 1.4 V, however if BAT falls below V<sub>SHUT</sub> after rising above V<sub>POR</sub>, the power on threshold depends on the minimum level reached by BAT after falling below V<sub>SHUT</sub>.

(2) Following POR, the device will operate down to this voltage.

## 7.7 3.3-V Voltage Regulator

| PARAMETER         |                                             | TEST CONDITION                                                      |  | MIN | TYP               | MAX  | UNIT |
|-------------------|---------------------------------------------|---------------------------------------------------------------------|--|-----|-------------------|------|------|
| V <sub>CTL</sub>  | Regulator control voltage <sup>(1)(2)</sup> | Measured at VCTL, V3P3 regulating                                   |  | 3.3 |                   | 26.4 | V    |
| V <sub>V3P3</sub> | Regulator output                            | Measured at V3P3, I <sub>REG</sub> = 0 to 4 mA, BAT = 4.2 to 26.4 V |  | 3.2 | 3.3               | 3.4  | V    |
| I <sub>REG</sub>  | V3P3 output current                         |                                                                     |  |     |                   | 4    | mA   |
| I <sub>SC</sub>   | V3P3 short circuit current limit            | V3P3 = 0.0 V                                                        |  | 10  |                   | 17   | mA   |
| V <sub>TB</sub>   | Thermistor bias voltage                     | Measured at VTb, I <sub>TB</sub> = 0                                |  |     | V <sub>V3P3</sub> |      | V    |
| I <sub>TB</sub>   | Thermistor bias current                     |                                                                     |  |     |                   | 1    | mA   |
| R <sub>TB</sub>   | Thermistor bias internal resistance         | R <sub>DS(ON)</sub> for internal FET switch, I <sub>TB</sub> = 1 mA |  |     | 90                | 130  | Ω    |

(1) When a bypass FET is used to supply the regulated 3.3 V load current, VCTL automatically adjusts to keep V3P3 = 3.3 V. Note that V<sub>CTL,MIN</sub> and the FET V<sub>GS</sub> will determine the minimum BAT voltage at which the bypass FET will operate.

(2) If VCTL is tied to BAT, the load current is supplied through V3P3.

## 7.8 Voltage Reference

| PARAMETER                                    | TEST CONDITION                                                    |                                                | MIN   | TYP                   | MAX   | UNIT                     |
|----------------------------------------------|-------------------------------------------------------------------|------------------------------------------------|-------|-----------------------|-------|--------------------------|
| $V_{REF}$ Voltage reference output           | Before gain correction,<br>$T_A = 25^\circ\text{C}$               | REF_SEL = 0                                    | 1.44  |                       | 1.56  | V                        |
|                                              |                                                                   | REF_SEL = 1                                    | 2.88  |                       | 3.12  |                          |
|                                              | After gain correction, <sup>(1)</sup><br>$T_A = 25^\circ\text{C}$ | REF_SEL = 0                                    | –0.1% | 1.5                   | +0.1% |                          |
|                                              |                                                                   | REF_SEL = 1                                    | –0.1% | 3                     | +0.1% |                          |
| $V_{REF\_CAL}$ Reference calibration voltage | Measured at V <sub>COUT</sub>                                     | V <sub>COUT_SEL</sub> = 2                      | –0.9% | $0.5 \times V_{REF}$  | +0.9% | V                        |
|                                              |                                                                   | V <sub>COUT_SEL</sub> = 3                      | –0.5% | $0.85 \times V_{REF}$ | +0.5% |                          |
|                                              |                                                                   | $(0.85 \times V_{REF}) - (0.5 \times V_{REF})$ | –0.3% | $0.35 \times V_{REF}$ | +0.3% | V                        |
| $\Delta V_{REF}$ Voltage reference tolerance | $T_A = 0 - 50^\circ\text{C}$                                      |                                                | –40   |                       |       | 40 ppm/ $^\circ\text{C}$ |
| $I_{REF}$ VREF output current                |                                                                   |                                                |       |                       |       | 10 $\mu\text{A}$         |

(1) Gain correction factor determined at final test and stored in non-volatile storage. Gain correction is applied by Host controller.

## 7.9 Cell Voltage Amplifier

| PARAMETER                                                             | TEST CONDITION                                                                                                                                      |                                                  | MIN   | TYP | MAX  | UNIT              |
|-----------------------------------------------------------------------|-----------------------------------------------------------------------------------------------------------------------------------------------------|--------------------------------------------------|-------|-----|------|-------------------|
| $G_{V_{COUT}}$ Cell voltage amplifier gain                            | Measured from VC <sub>n</sub> to V <sub>COUT</sub>                                                                                                  | REF_SEL = 0                                      | –1.6% | 0.3 | 1.5% |                   |
|                                                                       |                                                                                                                                                     | REF_SEL = 1                                      | –1.6% | 0.6 | 1.5% |                   |
| $O_{V_{COUT}}$ Cell voltage amplifier offset                          | Measured from VC <sub>n</sub> to V <sub>COUT</sub>                                                                                                  |                                                  | –16   |     |      | 15 mV             |
| $V_{COUT}$ Cell voltage amp output range <sup>(1)</sup>               | Measured at V <sub>COUT</sub> ,<br>VC <sub>n</sub> = 5 V                                                                                            | REF_SEL = 0                                      | 1.47  | 1.5 | 1.53 | V                 |
|                                                                       |                                                                                                                                                     | REF_SEL = 1                                      | 2.94  | 3   | 3.06 | V                 |
|                                                                       | Measured at V <sub>COUT</sub> ,<br>VC <sub>n</sub> = 0 V                                                                                            |                                                  | 0     |     |      | V                 |
| $\Delta V_{COUT}$ Cell voltage amplifier accuracy                     | VC <sub>n</sub> = 1.4 V to 4.4 V,<br>After correction, <sup>(2)</sup><br>Measured at V <sub>COUT</sub> <sup>(3)</sup><br>REF_SEL = 1 <sup>(4)</sup> | $T_A = 25^\circ\text{C}$                         | –3    |     |      | 3 mV              |
|                                                                       |                                                                                                                                                     | $T_A = 0^\circ\text{C}$ to $50^\circ\text{C}$    | –5    |     |      | 5                 |
|                                                                       |                                                                                                                                                     | $T_A = -25^\circ\text{C}$ to $+85^\circ\text{C}$ | –8    |     |      | 8                 |
|                                                                       |                                                                                                                                                     |                                                  |       |     |      |                   |
| $I_{V_{COUT}}$ V <sub>COUT</sub> output current <sup>(5)</sup>        |                                                                                                                                                     |                                                  |       |     |      | 10 $\mu\text{A}$  |
| $t_{V_{COUT}}$ Delay from VC <sub>n</sub> select to V <sub>COUT</sub> | Output step of 200 mV, C <sub>OUT</sub> = 0.1 $\mu\text{F}$                                                                                         |                                                  |       |     |      | 100 $\mu\text{s}$ |

(1) For VC<sub>n</sub> values greater than 5 V, V<sub>COUT</sub> clamps at approximately V<sub>3P3</sub>.

(2) Correction factor determined at final test and stored in non-volatile storage. Correction is applied by Host controller.

(3) Output referred. Input referred accuracy is calculated as  $\Delta V_{COUT} / G_{V_{COUT}}$  (for example, 3 / 0.6 = 5).

(4) Correction factors are calibrated for gain of 0.6. Tolerance at gain of 0.3 is approximately doubled. Contact TI for information on devices calibrated to a gain of 0.3.

(5) Max DC load for specified accuracy.

## 7.10 Current Sense Amplifier

| PARAMETER                     |                                                 | TEST CONDITION                                       |             | MIN  | TYP | MAX  | UNIT |
|-------------------------------|-------------------------------------------------|------------------------------------------------------|-------------|------|-----|------|------|
| G <sub>V<sub>IOUT</sub></sub> | Current sense amplifier gain                    | Measured from SENSEN,<br>SENSEP to V <sub>IOUT</sub> | I_GAIN = 0  | 4    |     |      |      |
|                               |                                                 |                                                      | I_GAIN = 1  | 8    |     |      |      |
| V <sub>IIN</sub>              | Current sense amp input range                   | Measured from SENSEN,<br>SENSEP to VSS               |             | −125 |     | 375  | mV   |
| V <sub>IOUT</sub>             | Current sense amp output range                  | Measured at V <sub>IOUT</sub>                        | REF_SEL = 0 | 0.25 |     | 1.25 | V    |
|                               |                                                 |                                                      | REF_SEL = 1 | 0.5  |     | 2.5  | V    |
|                               | Zero current output                             | Measured at V <sub>IOUT</sub><br>SENSEP = SENSEN     | REF_SEL = 0 |      | 1   |      | V    |
|                               |                                                 |                                                      | REF_SEL = 1 |      | 2   |      | V    |
| ΔV <sub>IOUT</sub>            | Current amplifier accuracy                      |                                                      |             | −1%  |     | 1%   |      |
| I <sub>V<sub>IOUT</sub></sub> | V <sub>IOUT</sub> output current <sup>(1)</sup> |                                                      |             |      |     | 10   | μA   |

(1) Max DC load for specified accuracy

## 7.11 Overcurrent Comparator

| PARAMETER                   | TEST CONDITION                                       | MIN                                      | TYP | MAX  | UNIT  |
|-----------------------------|------------------------------------------------------|------------------------------------------|-----|------|-------|
| V <sub>BAT_COMP</sub>       | Minimum VBAT for comparator operation <sup>(1)</sup> |                                          |     | 5    | V     |
| G <sub>VCOMP</sub>          | Comparator amplifier gain                            | Measured from SENSEP to comparator input |     | 1    |       |
| V <sub>ITRIP</sub>          | Current comparator trip threshold <sup>(2)</sup>     | 25                                       |     | 400  | mV    |
| $\Delta$ V <sub>ITRIP</sub> | Current comparator accuracy                          | V <sub>ITRIP</sub> = 25 mV               |     | –6   | 6 mV  |
|                             |                                                      | V <sub>ITRIP</sub> > 25 mV               |     | –10% | 10% V |
| V <sub>OL_ALERT</sub>       | ALERT Output Low Logic                               | I <sub>ALERT</sub> = 1 mA                |     | 0.4  | V     |
| V <sub>OH_ALERT</sub>       | ALERT Output High Logic <sup>(3)</sup>               | NA                                       | NA  | NA   |       |
| I <sub>ALERT</sub>          | ALERT Pulldown current                               | ALERT = 0.4 V, Output driving low        |     | 1    | mA    |
| I <sub>ALERT_LKG</sub>      | ALERT Leakage current                                | ALERT = 5 V, Output Hi-Z                 |     | < 1  | μA    |
| t <sub>OC</sub>             | Comparator response time                             |                                          |     | 100  | μs    |

(1) The Overcurrent Comparator is not guaranteed to work when VBAT is below this voltage.

(2) Trip threshold selectable from 25, 50, 75, 100, 125, 150, 175, 200, 225, 250, 275, 300, 325, 350, 375 or 400 mV.

(3) This parameter NA because output is open drain.

## 7.12 Internal Temperature Measurement

| PARAMETER                      | TEST CONDITION                           | MIN                                                     | TYP  | MAX  | UNIT       |
|--------------------------------|------------------------------------------|---------------------------------------------------------|------|------|------------|
| V <sub>TEMP_INT</sub>          | Internal temperature voltage             | Measured at V <sub>COUT</sub> , T <sub>INT</sub> = 25°C |      | 1.15 | 1.2 1.25 V |
| $\Delta$ V <sub>TEMP_INT</sub> | Internal temperature voltage sensitivity |                                                         | –4.4 |      | mV/°C      |

## 7.13 Cell Balancing and Open Cell Detection

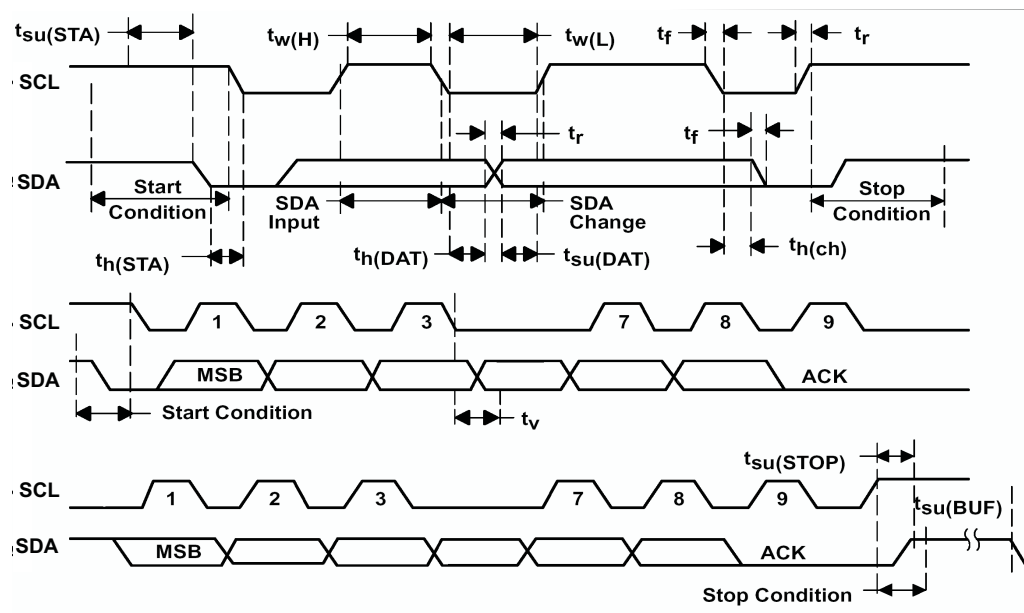
| PARAMETER        | TEST CONDITION                                                                  | MIN | TYP | MAX | UNIT |
|------------------|---------------------------------------------------------------------------------|-----|-----|-----|------|
| R <sub>BAL</sub> | R <sub>DS(ON)</sub> for VC1 internal FET switch, VC <sub>n</sub> = 3.6 V        | 1   | 3   | 5   | Ω    |
|                  | R <sub>DS(ON)</sub> for internal VC2 to VC6 FET switch, VC <sub>n</sub> = 3.6 V | 3   | 5.5 | 8   |      |

(1) Balancing current is not internally limited. The cell balancing operation is completely controlled by the Host processor, no automatic function or time-out is included in the part. Take care to ensure that balancing current through the part is below the maximum power dissipation limit. The Host algorithm is responsible for limiting thermal dissipation to package ratings.

## 7.14 I<sup>2</sup>C Compatible Interface

| PARAMETERS            |                                                                      |                           | MIN              | TYP | MAX  | UNIT |
|-----------------------|----------------------------------------------------------------------|---------------------------|------------------|-----|------|------|
| <b>DC PARAMETERS</b>  |                                                                      |                           |                  |     |      |      |
| V <sub>IL</sub>       | Input Low Logic Threshold                                            |                           |                  |     | 0.6  | V    |
| V <sub>IH</sub>       | Input High Logic Threshold                                           |                           | 2.8              |     |      | V    |
| V <sub>OL</sub>       | Output Low Logic Drive                                               | I <sub>OL</sub> = 1 mA    |                  |     | 0.20 | V    |
|                       |                                                                      | I <sub>OL</sub> = 2.5 mA  |                  |     | 0.40 |      |
| V <sub>OH</sub>       | Output High Logic Drive (Not applicable due to open-drain outputs)   |                           |                  | N/A |      | V    |
| I <sub>LKG</sub>      | I <sup>2</sup> C Pin Leakage                                         | Pin = 5 V, Output in Hi-Z |                  |     | < 1  | μA   |
| <b>AC PARAMETERS</b>  |                                                                      |                           |                  |     |      |      |
| t <sub>r</sub>        | SCL, SDA Rise Time                                                   |                           |                  |     | 1000 | ns   |
| t <sub>f</sub>        | SCL, SDA Fall Time                                                   |                           |                  |     | 300  | ns   |
| t <sub>w(H)</sub>     | SCL Pulse Width High                                                 |                           | 4                |     |      | μs   |
| t <sub>w(L)</sub>     | SCL Pulse Width Low                                                  |                           | 4.7              |     |      | μs   |
| t <sub>su(STA)</sub>  | Setup time for START condition                                       |                           | 4.7              |     |      | μs   |
| t <sub>h(STA)</sub>   | START condition hold time after which first clock pulse is generated |                           | 4                |     |      | μs   |
| t <sub>su(DAT)</sub>  | Data setup time                                                      |                           | 250              |     |      | ns   |
| t <sub>h(DAT)</sub>   | Data hold time                                                       |                           | 0 <sup>(1)</sup> |     |      | μs   |
| t <sub>su(STOP)</sub> | Setup time for STOP condition                                        |                           | 4                |     |      | μs   |
| t <sub>su(BUF)</sub>  | Time the bus must be free before new transmission can start          |                           | 4.7              |     |      | μs   |
| t <sub>v</sub>        | Clock Low to Data Out Valid                                          |                           |                  |     | 900  | ns   |
| t <sub>h(CH)</sub>    | Data Out Hold Time After Clock Low                                   |                           | 0                |     |      | ns   |
| f <sub>SCL</sub>      | Clock Frequency                                                      |                           | 0                |     | 100  | kHz  |
| t <sub>WAKE</sub>     | I <sup>2</sup> C ready after transition to Wake Mode                 |                           |                  |     | 2.5  | ms   |

- (1) Devices must provide internal hold time of at least 300 ns for the SDA signal-to-bridge of the undefined region of the falling edge of SCL.



**Figure 1. I<sup>2</sup>C Timing**

## 7.15 Typical Characteristics

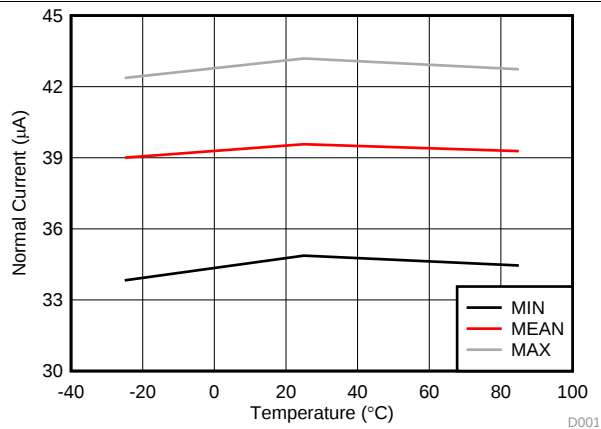


Figure 2. Normal Mode Supply Current

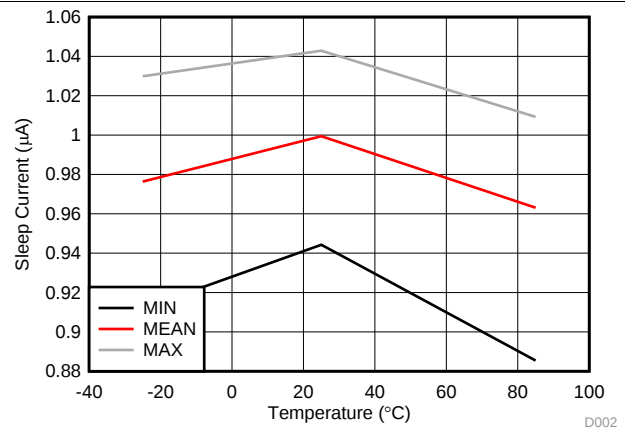


Figure 3. Sleep Mode Supply Current

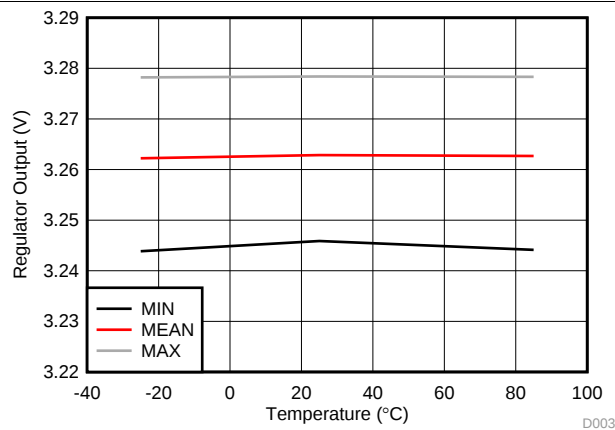


Figure 4. Regulator Output With 4 mA Load

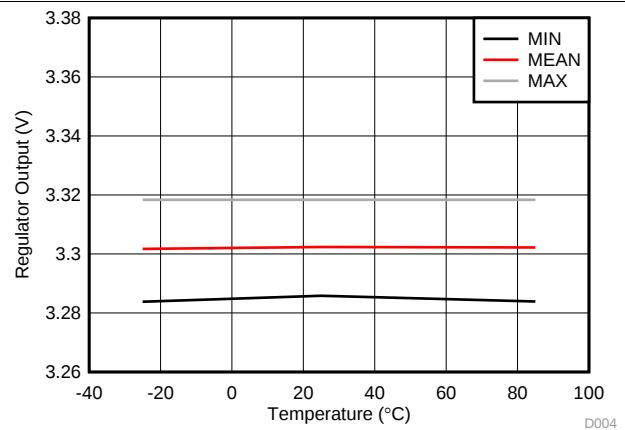


Figure 5. Regulator Output With No Load

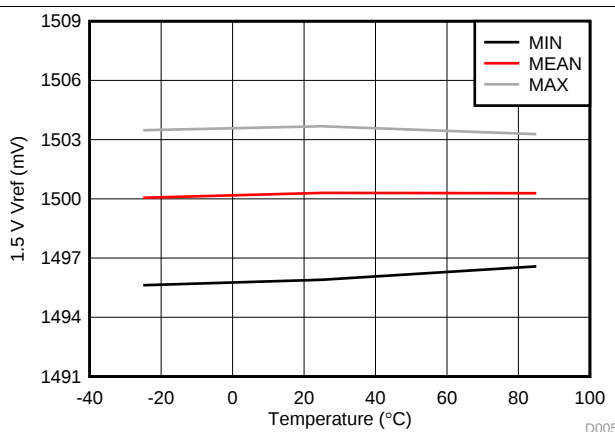


Figure 6. 1.5-V VREF Output (Before Correction)

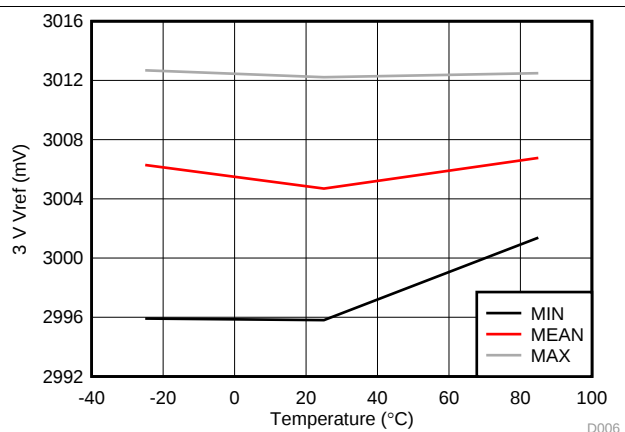


Figure 7. 3-V VREF Output (Before Correction)

## 8 Detailed Description

### 8.1 Overview

The bq76925 Host-controlled analog front end (AFE) is part of a complete pack monitoring, balancing, and protection system for 3-series to 6-series cell Lithium batteries. The bq76925 allows a Host controller to easily monitor individual cell voltages, pack current, and temperature. The Host may use this information to detect and act on a fault condition caused when one or more of these parameters exceed the limits of the application. In addition, the Host may use this information to determine end-of-charge, end-of-discharge, and other gas-gauging and state of health conditions.

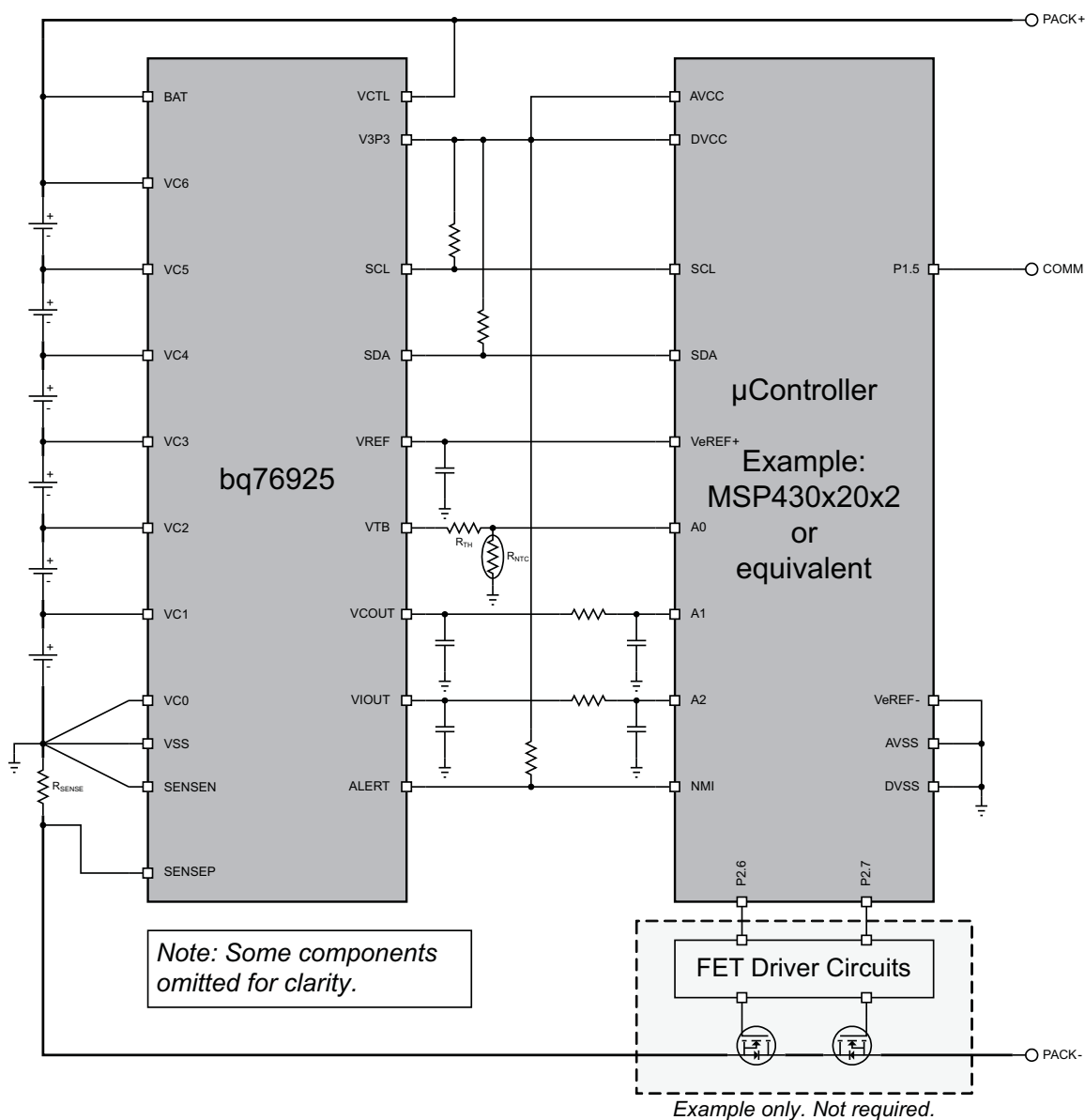
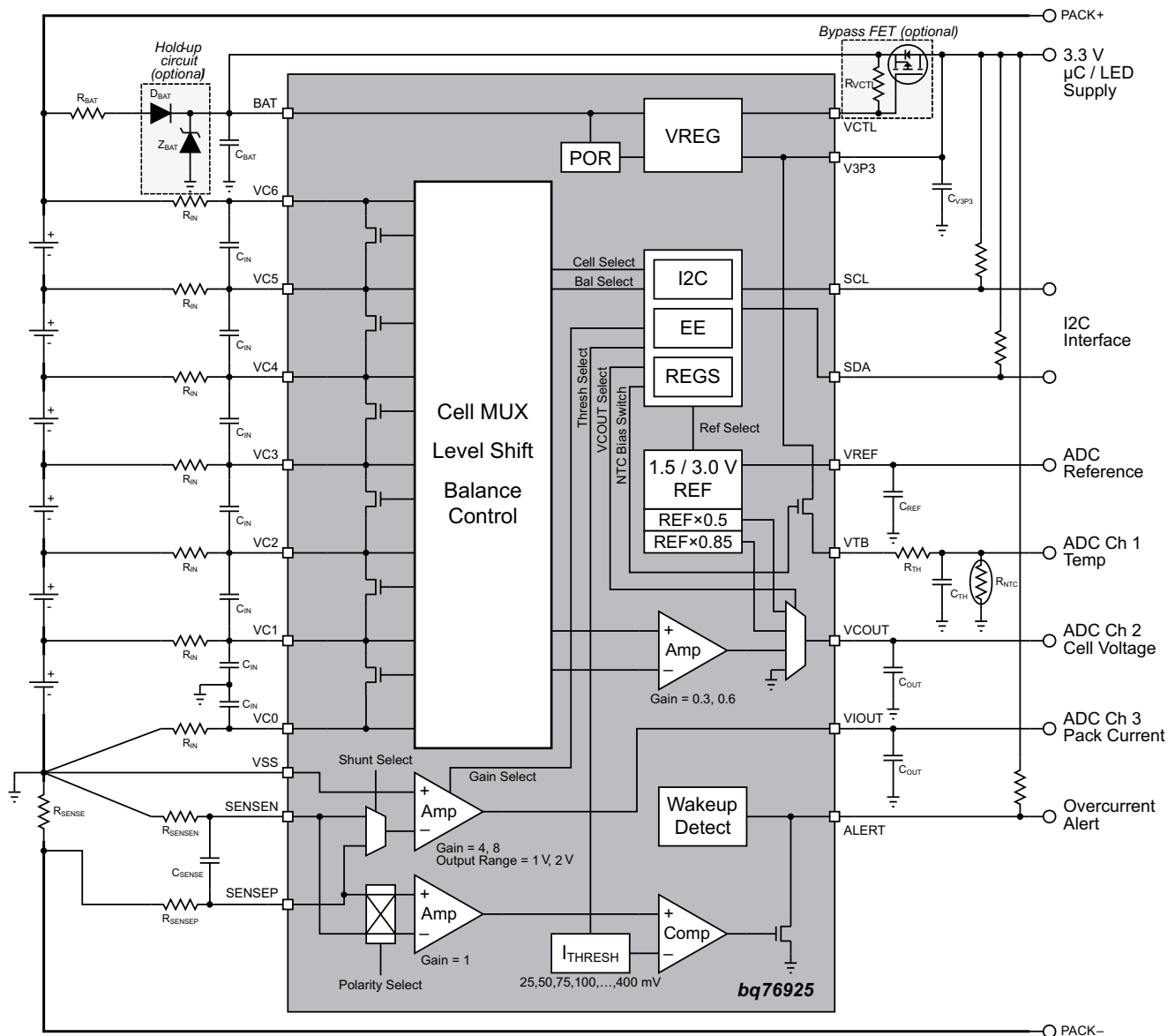


Figure 8. Example of bq76925 With Host Controller

## 8.2 Functional Block Diagram

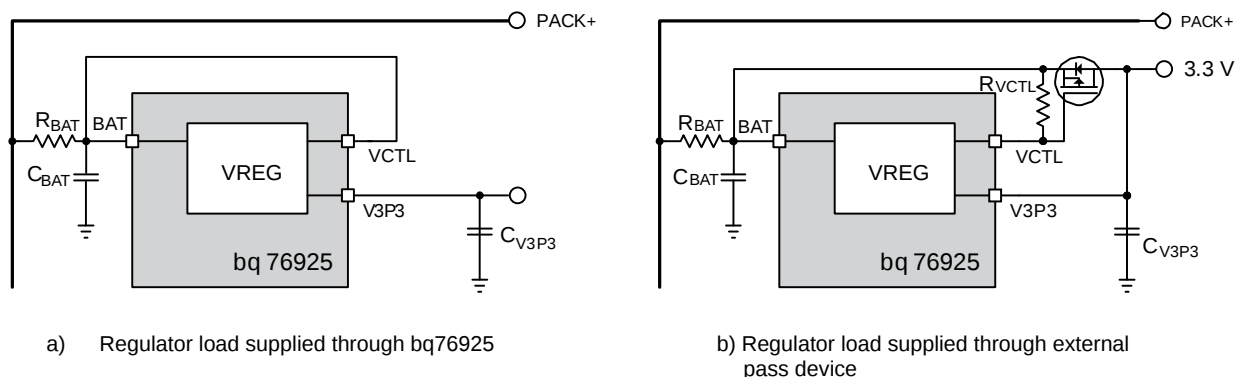


## 8.3 Feature Description

### 8.3.1 Internal LDO Voltage Regulator

The bq76925 device provides a regulated 3.3-V supply voltage on the V3P3 pin for operating the device's internal logic and interface circuitry. This regulator may also be used to directly power an external microcontroller or other external circuitry up to a limit of 4-mA load current. In this configuration, the VCTL pin is tied directly to the BAT pin. For applications requiring more than 4 mA, an external-bypass transistor may be used to supply the load current. In this configuration, the VCTL pin is tied to the gate of the bypass FET. These two configurations are shown in [Figure 9](#).

## Feature Description (continued)



**Figure 9. LDO Regulator Configurations**

For the configuration of [Figure 9b](#)), a high-gain bypass device should be used to ensure stability. A bipolar PNP or p-channel FET bypass device may be used. Contact TI for recommendations.

The LDO regulator may be overridden (that is, back-fed) by an external-supply voltage greater than the regulated voltage on V3P3. In this configuration, the bq76925 internal logic and interface circuitry operates from the external supply and the internal 3.3-V regulator supplies no load current.

### 8.3.2 ADC Interface

The bq76925 device is designed to interface to a multi-channel analog-to-digital converter (ADC) located in an external Host controller, such as an MSP430 Microcontroller or equivalent. Three outputs provide voltage, current, and temperature information for measurement by the Host. In addition, the bq76925 device includes a low-drift calibrated 1.5 / 3 V reference that is output on a dedicated pin for use as the reference input to the ADC.

The gain and offset characteristics of the bq76925 device are measured during factory test and stored in non-volatile memory as correction factors. The Host reads these correction factors and applies them to the ADC conversion results in order to achieve high-measurement accuracy. In addition, the precise voltage reference of the bq76925 can be used to calibrate the gain and offset of the Host ADC.

#### 8.3.2.1 Reference Voltage

The bq76925 device outputs a stable reference voltage for use by the Host ADC. A nominal voltage of 1.5 V or 3 V is selected through the REF\_SEL bit in the CONFIG\_2 register. The reference voltage is very stable across temperature, but the initial voltage may vary by  $\pm 4\%$ . The variation from nominal is manifested as a gain error in the ADC conversion result. To correct for this error, offset and gain correction factors are determined at final test and stored in the non-volatile registers VREF\_CAL and VREF\_CAL\_EXT. The Host reads the correction factors and applies them to the nominal reference voltage to arrive at the actual reference voltage as described under [Cell Voltage Monitoring](#). After gain correction, the tolerance of the reference will be within  $\pm 0.1\%$ .

##### 8.3.2.1.1 Host ADC Calibration

All analog-to-digital converters have inherent gain and offset errors, which adversely affect measurement accuracy. Some microcontrollers may be characterized by the manufacturer and shipped with ADC gain and offset information stored on-chip. It is also possible for such characterization to be done by the end-user on loose devices prior to PCB assembly or as a part of the assembled PCB test.

For applications where such ADC characterization is not provided or is not practical, the bq76925 device provides a means for in-situ calibration of the Host ADC through setting of the VCOUT\_SEL bits in the CELL\_CTL register two scaled versions of the reference voltage,  $0.5 \times V_{REF}$  and  $0.85 \times V_{REF}$ , can be selected for output on the VCOUT pin for measurement by the Host ADC. Measuring both scaled voltages enables the Host to do a two-point calibration of the ADC and compensate for the ADC offset and gain in all subsequent ADC measurement results as shown in [Figure 10](#).

## Feature Description (continued)

Note that the calibration accuracy will be limited by the tolerance of the scaled reference-voltage output so that use of this method may not be effective. For these cases, TI recommends to use a higher-accuracy source for the two-point calibration shown in Figure 10.

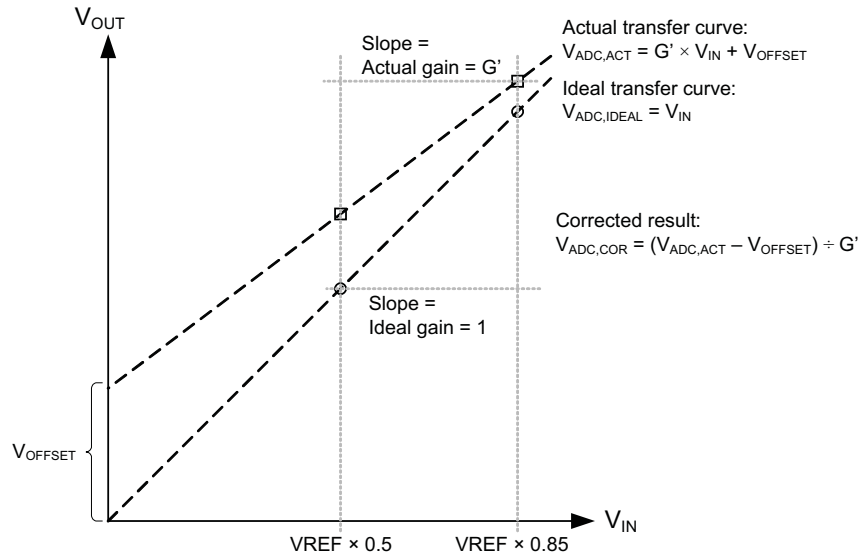


Figure 10. Host ADC Calibration Using  $V_{REF}$

### 8.3.2.2 Cell Voltage Monitoring

The cell-voltage monitoring circuits include an input level-shifter, multiplexer (MUX), and scaling amplifier. The Host selects one  $VC_n$  cell input for measurement by setting the  $VCOUT\_SEL$  and  $CELL\_SEL$  bits in the  $CELL\_CTL$  register. The scaling factor is set by the  $REF\_SEL$  bit in the  $CONFIG\_2$  register. The selected cell input is level shifted to  $V_{SS}$  reference, scaled by a nominal gain  $G_{VCOUT} = 0.3$  ( $REF\_SEL = 0$ ) or  $0.6$  ( $REF\_SEL = 1$ ) and output on the  $VCOUT$  pin for measurement by the Host ADC.

Similar to the reference voltage, gain and offset correction factors are determined at final test for each individual cell input and stored in non-volatile registers  $VC_n\_CAL$  ( $n = 1-6$ ) and  $VC\_CAL\_EXT\_m$  ( $m = 1-2$ ). These factors are read by the Host and applied to the ADC voltage-measurement results in order to obtain the specified accuracy.

The cell voltage offset and gain correction factors are stored as 5-bit signed integers in 2's complement format. The most significant bits ( $VC_n\_OC\_4$ ,  $VC_n\_GC\_4$ ) are stored separately and must be concatenated with the least significant bits ( $VC_n\_OFFSET\_CORR$ ,  $VC_n\_GAIN\_CORR$ ).

The reference voltage offset and gain correction factors are stored respectively as a 6-bit and 5-bit signed integer in 2's complement format. As with the cell voltage correction factors, the most significant bits ( $V_{REF\_OC\_5}$ ,  $V_{REF\_OC\_4}$ ,  $V_{REF\_GC\_4}$ ) are stored separately and must be concatenated with the least significant bits ( $V_{REF\_OFFSET\_CORR}$ ,  $V_{REF\_GAIN\_CORR}$ ).

The actual cell voltage ( $VC_n$ ) is calculated from the measured voltage ( $VCOUT$ ) as shown in the following equations:

$$VCOUT = \frac{\text{ADC Count}}{\text{Full Scale Count}} \times V_{REF\_NOMINAL}$$

$$VC_n = \frac{VCOUT \times GC_{V_{REF}} + OC_{VCOUT}}{G_{VCOUT}} \times (1 + GC_{VCOUT}) \quad (1)$$

## Feature Description (continued)

$$GC_{VCOUT} = \left[ (VCn\_GC\_4 \ll 4) + VCn\_GAIN\_CORR \right] \times 0.001,$$

$$OC_{VCOUT} = \left[ (VCn\_OC\_4 \ll 4) + VCn\_OFFSET\_CORR \right] \times 0.001,$$

$$GC_{VREF} = (1 + \left[ (VREF\_GC\_4 \ll 4) + VREF\_GAIN\_CORR \right] \times 0.001) + \frac{\left[ (VREF\_OC\_5 \ll 5) + (VREF\_OC\_4 \ll 4) + VREF\_OFFSET\_CORR \right] \times 0.001}{VREF_{NOMINAL}}$$
(2)

### 8.3.2.2.1 Cell Amplifier Headroom Under Extreme Cell Imbalance

For cell voltages across  $(VC1 - VC0)$  that are less than approximately 2.64 V, extreme cell-voltage imbalances between  $(VC1 - VC0)$  and  $(VC2 - VC1)$  can lead to a loss of gain in the  $(VC2 - VC1)$  amplifier. The cell imbalance at which the loss of gain occurs is determined by [Equation 3](#):

$$(VC2 - VC1) \times 0.6 > (VC1 - VSS)$$
(3)

Assuming  $VC0 = VSS$ , it can be seen that when  $(VC1 - VC0) > 2.64$  volts, the voltage across  $(VC2 - VC1)$  can range up to the limit of 4.4 V without any loss of gain. At the minimum value of  $(VC1 - VC0) = 1.4$  V, an imbalance of more than 900 mV is tolerated before any loss of gain in the  $(VC2 - VC1)$  amplifier. For higher values of  $(VC1 - VC0)$ , increasingly large imbalances are tolerated. For example, when  $(VC1 - VC0) = 2.0$  V, an imbalance up to 1.33 V (that is,  $(VC2 - VC1) = 3.33$  V) results in no degradation of amplifier performance.

Normally, cell imbalances greater than 900 mV will signal a faulty condition of the battery pack and its use should be discontinued. *The loss of gain on the second cell input does not affect the ability of the system to detect this condition. The gain fall-off is gradual so that the measured imbalance will never be less than the critical imbalance set by [Equation 3](#).*

Therefore, if the measured  $(VC2 - VC1)$  is greater than  $(VC1 - VSS) / 0.6$ , a severe imbalance is detected and the pack should enter a fault state which prevents further use. In this severe cell imbalance condition comparisons of the measured  $(VC2 - VC1)$  to any overvoltage limits will be optimistic due to the reduced gain in the amplifier, further emphasizing the need to enter a fault state.

### 8.3.2.2.2 Cell Amplifier Headroom Under BAT Voltage Drop

Voltage differences between BAT and the top cell potential come from two sources as shown in [Figure 11](#): V3P3 regulator current that flows through the  $R_{BAT}$  filter resistor, and the voltage drop in the series diode  $D_{BAT}$  of the hold-up circuit. These effects cause BAT to be less than the top-cell voltage measured by the cell amplifier.

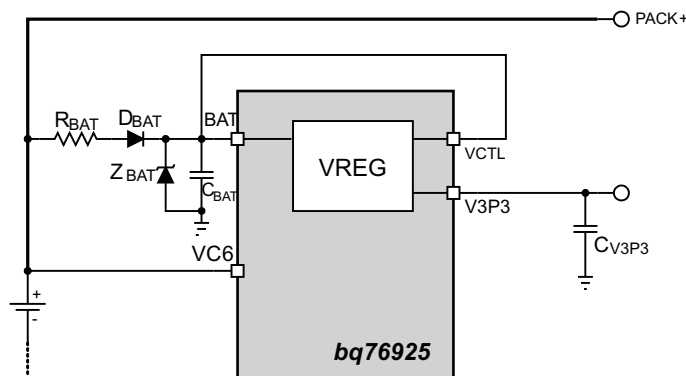


Figure 11. Sources of Voltage Drop Affecting the BAT Pin

## Feature Description (continued)

The top-cell amplifier (VC6 – VC5) is designed to measure an input voltage down to 1.4 V with a difference between the BAT and VC6 pin up to 1.2 V (that is, BAT can be 1.2 V lower than VC6). However, in applications with fewer than 6 cells, the upper-cell inputs are typically shorted to the top-cell input. For example, in a 5-cell application VC6 and VC5 would be shorted together and the (VC5 – VC4) amplifier would measure the top-cell voltage. The case is similar for 4-cell and 3-cell applications.

For these cases when using the (VC5 – VC4), (VC4 – VC3), or (VC3 – VC2) amplifier to measure the top cell, the difference between BAT and the top-cell amplifier must be less than 240 mV in order to measure cell voltages down to 1.4 V. Note that at higher-cell input voltages the top amplifier tolerates a greater difference. For example, in a 5-cell configuration (VC6 and VC5 tied together) the (VC5 – VC4) amplifier is able to measure down to a 1.7 V input with a 600-mV difference between VC5 and BAT.

Accordingly, in systems with fewer than 6 cells, it is important in system design to minimize  $R_{BAT}$  and to use a Schottky type diode for  $D_{BAT}$  with a low forward voltage. If it is not possible to reduce the drop at BAT to an acceptable level, then for 4-cell and 5-cell configurations, the (VC6 – VC5) amplifier may be used as the top cell amplifier as shown in [Table 1](#), which allows up to a 1.2 V difference between BAT and the top cell.

**Table 1. Alternate Connections for 4 and 5 Cells**

| Configuration | Cell 5    | Cell 4    | Cell 3    | Cell 2    | Cell 1    | Unused Cell Inputs      |
|---------------|-----------|-----------|-----------|-----------|-----------|-------------------------|
| 5-cell        | VC6 – VC5 | VC4 – VC3 | VC3 – VC2 | VC2 – VC1 | VC1 – VC0 | Short VC5 to VC4        |
| 4-cell        |           | VC6 – VC5 | VC3 – VC2 | VC2 – VC1 | VC1 – VC0 | Short VC5 to VC4 to VC3 |

### 8.3.2.3 Current Monitoring

Current is measured by converting current to voltage through a sense resistor connected between SENSEN and SENSEP. A positive voltage at SENSEP with respect to SENSEN indicates a discharge current is flowing, and a negative voltage indicates a charge current. The small voltage developed across the sense resistor is amplified by gain  $G_{VIOUT}$  and output on the VIOUT pin for conversion by the Host ADC. The voltage on VIOUT is always positive and for zero current is set to 3/4 of the output range. The current sense amplifier is inverting; discharge current causes VIOUT to decrease and charge current causes VIOUT to increase. Therefore, the measurement range for discharge currents is 3 times the measurement range for charge currents.

The current-sense amplifier is preceded by a multiplexer that allows measurement of either the SENSEN or SENSEP input with respect to VSS. The Host selects the pin for measurement by writing the  $I\_AMP\_CAL$  bit in the  $CONFIG\_1$  register. The Host then calculates the voltage across the sense resistor by subtracting the measured voltage at SENSEN from the measured voltage at SENSEP. If the SENSEN and VSS connections are such that charge and discharge currents do not flow through the connection between them; that is, there is no voltage drop between SENSEN and VSS due to the current being measured, then the measurement of the SENSEN voltage can be regarded as a calibration step and stored by the Host for use as a pseudo-constant in the  $V_{SENSE}$  calculation. The SENSEN voltage measurement would then only need updating when changing environmental conditions warrant.

The Host sets  $G_{VIOUT}$  by writing the  $I\_GAIN$  bit in the  $CONFIG\_1$  register. The available gains of 4 and 8 enable operation with a variety of sense-resistor values over a broad range of pack currents. The gain may be changed at any time allowing for dynamic range and resolution adjustment. The input and output ranges of the amplifier are determined by the value of the  $REF\_SEL$  bit in the  $CONFIG\_2$  register. These values are shown in [Table 2](#). Because the current amplifier is inverting, the Min column under Output Range corresponds to the Max column under Input Range. Likewise, the Max column under Output Range corresponds to the Min column under Input Range.

The actual current is calculated from the measured voltage (VIOUT) as follows. Note that  $V_{SENSE}$  is positive when discharge current is flowing. In keeping with battery pack conventions, the sign of  $I_{SENSE}$  is inverted so that discharge current is negative.

$$V_{SENSE} = \frac{-(VIOUT(SENSEP) - VIOUT(SENSEN))}{G_{VIOUT}}$$

$$I_{SENSE} = -\frac{V_{SENSE}}{R_{SENSE}} \quad (4)$$

**Table 2. Current Amplifier Configurations**

| REF_SEL | I_GAIN | Gain | V <sub>IOUT</sub> (V) at<br>I <sub>SENSE</sub> = 0<br>(typical) | Input Range <sup>(1)</sup><br>(mV) |       | Output Range <sup>(2)</sup><br>(V) |      | I <sub>SENSE</sub> Range (A) at<br>R <sub>SENSE</sub> = 1 mΩ | I <sub>SENSE</sub> Resolution<br>(mA)/10-bit<br>ADC <sup>(3)</sup> |
|---------|--------|------|-----------------------------------------------------------------|------------------------------------|-------|------------------------------------|------|--------------------------------------------------------------|--------------------------------------------------------------------|
|         |        |      |                                                                 | Min                                | Max   | Min                                | Max  |                                                              |                                                                    |
| 0       | 0      | 4    | 1.0                                                             | –62.5                              | 187.5 | 0.25                               | 1.25 | –62.5 – 187.5                                                | 366                                                                |
| 0       | 1      | 8    | 1.0                                                             | –14                                | 91    | 0.27                               | 1.11 | –14 – 91                                                     | 183                                                                |
| 1       | 0      | 4    | 2.0                                                             | –125                               | 375   | 0.5                                | 2.5  | –125 – 375                                                   | 732                                                                |
| 1       | 1      | 8    | 2.0                                                             | –62.5                              | 187.5 | 0.5                                | 2.5  | –62.5 – 187.5                                                | 366                                                                |

(1) SENSEN or SENSEP measured with respect to VSS.

(2) Output range assumes typical value of V<sub>IOUT</sub> at I<sub>SENSE</sub> = 0. For non-typical values, the output range will shift accordingly.

(3) Assumes 1 mΩ R<sub>SENSE</sub> and ADC reference voltage of 1.5 V and 3.0 V when REF\_SEL = 0 and 1, respectively.

### 8.3.2.4 Overcurrent Monitoring

The bq76925 device also includes a comparator for monitoring the current-sense resistor and alerting the Host when the voltage across the sense resistor exceeds a selected threshold. The available thresholds range from 25 mV to 400 mV and are set by writing the I\_THRESH bits in the CONFIG\_1 register. Positive (discharge) or negative (charge) current may be monitored by setting the I\_COMP\_POL bit in the CONFIG\_1 register. By the choice of sense resistor and threshold, a variety of trip points are possible to support a wide range of applications.

The comparator result is driven through the open-drain ALERT output to signal the host when the threshold is exceeded. This feature can be used to wake up the Host on connection of a load or to alert the Host to a potential fault condition. The ALERT pin state is also available by reading the ALERT bit in the STATUS register.

### 8.3.2.5 Temperature Monitoring

To enable temperature measurements by the Host, the bq76925 device provides the LDO regulator voltage on a separate output pin (VTB) for biasing an external thermistor network. In order to minimize power consumption, the Host may switch the VTB output on and off by writing to the VTB\_EN bit in the POWER\_CTL register. Note that if the LDO is back-fed by an external source, the VTB bias will be switched to the external source.

In a typical application, the thermistor network will consist of a resistor in series with an NTC thermistor, forming a resistor divider where the output is proportional to temperature. This output may be measured by the Host ADC to determine temperature.

#### 8.3.2.5.1 Internal Temperature Monitoring

The internal temperature (T<sub>INT</sub>) of the bq76925 device can be measured by setting V<sub>COUT\_SEL</sub> = '01' and CELL\_SEL = '110' in the CELL\_CTL register. In this configuration, a voltage proportional to temperature (V<sub>TEMP\_INT</sub>) is output on the V<sub>COUT</sub> pin. This voltage is related to the internal temperature as follows:

$$V_{TEMP\_INT}(mV) = V_{TEMP\_INT}(T_{INT} = 25^{\circ}C) - T_{INT}(^{\circ}C) \times \Delta V_{TEMP\_INT} \quad (5)$$

### 8.3.3 Cell Balancing and Open Cell Detection

The bq76925 device integrates cell-balancing FETs that are individually controlled by the Host. The balancing method is resistive bleed balancing, where the balancing current is set by the external cell input resistors. The maximum allowed balancing current is 50 mA per cell.

The Host may activate one or more cell balancing FETs by writing the BAL\_n bits in the BAL\_CTL register. To allow the greatest flexibility, the Host has complete control over the balancing FETs. However, in order to avoid exceeding the maximum cell input voltage, the bq76925 will prevent two adjacent balancing FETs from being turned on simultaneously. If two adjacent bits in the balance control register are set to 1, neither balancing transistor will be turned on. The Host based balancing algorithm must also limit the power dissipation to the maximum ratings of the device.

In a normal system, closing a cell-balancing FET will cause 2 cell voltages to appear across one cell input. This fact can be utilized to detect a cell sense-line open condition, that is, a broken wire from the cell-sense point to the bq76925 VC<sub>n</sub> input. Table 3 shows how this can be accomplished. Note that the normal cell-voltage measurements may represent a saturated or full-scale reading. However, these will normally be distinguishable from the open-cell measurement.

**Table 3. Open Cell Detection Method**

| Kelvin input to test | Method 1 |         |                            |       | Method 2 |         |                            |       |
|----------------------|----------|---------|----------------------------|-------|----------|---------|----------------------------|-------|
|                      | Turn On  | Measure | Result                     |       | Turn On  | Measure | Result                     |       |
|                      |          |         | Normal                     | Open  |          |         | Normal                     | Open  |
| VC0                  | BAL_1    | CELL2   | $CELL2 + 0.5 \times CELL1$ | CELL2 |          |         |                            |       |
| VC1                  | BAL_2    | CELL3   | $CELL3 + 0.5 \times CELL2$ | CELL3 |          |         |                            |       |
| VC2                  | BAL_3    | CELL4   | $CELL4 + 0.5 \times CELL3$ | CELL4 | BAL_2    | CELL1   | $CELL1 + 0.5 \times CELL2$ | CELL1 |
| VC3                  | BAL_4    | CELL5   | $CELL5 + 0.5 \times CELL4$ | CELL5 | BAL_3    | CELL2   | $CELL2 + 0.5 \times CELL3$ | CELL2 |
| VC4                  | BAL_5    | CELL6   | $CELL6 + 0.5 \times CELL5$ | CELL6 | BAL_4    | CELL3   | $CELL3 + 0.5 \times CELL4$ | CELL3 |
| VC5                  |          |         |                            |       | BAL_5    | CELL4   | $CELL4 + 0.5 \times CELL5$ | CELL4 |
| VC6                  |          |         |                            |       | BAL_6    | CELL5   | $CELL5 + 0.5 \times CELL6$ | CELL5 |

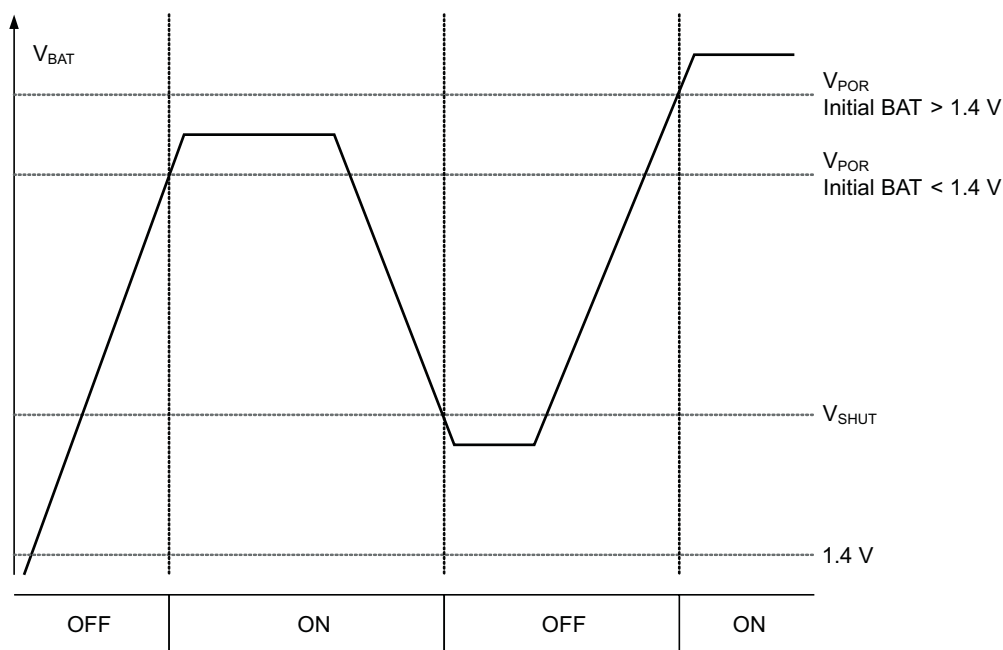
Note that the cell amplifier headroom limits discussed above apply to the open-cell detection method because by virtue of closing a switch between 2 cell inputs, internal to the device this appears as an extreme cell imbalance. Therefore, when testing for an open on CELL2 by closing the CELL1 balancing FET, the CELL2 measurement will be less than the expected normal result due to gain loss caused by the imbalance. However, the CELL2 measurement will still increase under this condition so that a difference between open (no change) and normal (measured voltage increases) can be detected.

## 8.4 Device Functional Modes

### 8.4.1 Power Modes

#### 8.4.1.1 POWER ON RESET (POR)

When initially powering up the bq76925 device, the voltage on the BAT pin must exceed  $V_{POR}$  (4.7-V maximum) before the device will turn on. Following this, the device will remain operational as long as the voltage on BAT remains above  $V_{SHUT}$  (3.6-V maximum). If the BAT voltage falls below  $V_{SHUT}$ , the device will shut down. Recovery from shutdown occurs when BAT rises back above the  $V_{POR}$  threshold and is equivalent to a POR. The  $V_{POR}$  threshold following a shutdown depends on the minimum level reached by BAT after crossing below  $V_{SHUT}$ . If BAT does not fall below approximately 1.4 V, a higher  $V_{POR}$  (7.5-V maximum) applies. This is illustrated in Figure 12.


**Figure 12. Power On State vs  $V_{BAT}$**

## Device Functional Modes (continued)

Following a power on reset, all volatile registers assume their default state. Therefore, care must be taken that transients on the BAT pin during normal operation do not fall below  $V_{SHUT}$ . To avoid this condition in systems subject to extreme transients or brown-outs, a hold-up circuit such as the one shown in the functional diagram is recommended. When using a hold-up circuit, care must be taken to observe the BAT to VC6 maximum ratings.

### 8.4.1.2 STANDBY

Individual device functions such as cell translator, current amplifier, reference, and current comparator can be enabled and disabled under Host control by writing to the POWER\_CTL register. The STANDBY feature can be used to save power by disabling functions that are unused. In the minimum power standby mode, all device functions can be turned off leaving only the 3.3-V regulator active.

### 8.4.1.3 SLEEP

In addition to STANDBY, there is also a SLEEP mode. In SLEEP mode the Host orders the bq76925 device to shutdown all internal circuitry and all functions including the LDO regulator. The device consumes a minimal amount of current ( $< 1.5 \mu A$ ) in SLEEP mode due only to leakage and powering of the wake-up detection circuitry.

SLEEP mode is entered by writing a '1' to the SLEEP bit in the POWER\_CTL register. Wake-up is achieved by pulling up the ALERT pin; however, the wake-up circuitry is not armed until the voltage at V3P3 drops to approximately 0 V. To facilitate the discharge of V3P3, an internal 3-k $\Omega$  pulldown resistor is connected from V3P3 to VSS during the time that sleep mode is active. Once V3P3 is discharged, the bq76925 may be awakened by pulling the ALERT pin above  $V_{WAKE}$  (2-V maximum).

The SLEEP\_DIS bit in the POWER\_CTL register acts as an override to the SLEEP function. When SLEEP\_DIS is set to '1', writing the SLEEP bit has no effect (that is, SLEEP mode cannot be entered). If SLEEP\_DIS is set after SLEEP mode has been entered, the device will immediately exit SLEEP mode. This scenario can arise if SLEEP\_DIS is set after SLEEP is set, but before V3P3 has discharged below a valid operating voltage. This scenario can also occur if the V3P3 pin is held up by external circuitry and not allowed to fully discharge.

If the overcurrent alert function is not used, the ALERT pin can function as a dedicated wake-up pin. Otherwise, the ALERT pin will normally be pulled up to the LDO voltage, so care must be taken in the system design so that the wake-up signal does not interfere with proper operation of the regulator.

## 8.5 Programming

### 8.5.1 Host Interface

The Host communicates with the AFE through an I<sup>2</sup>C interface. A CRC byte may optionally be used to ensure robust operation. The CRC is calculated over all bytes in the message according to the polynomial  $x^8 + x^2 + x + 1$ .

#### 8.5.1.1 I<sup>2</sup>C Addressing

In order to reduce communications overhead, the addressing scheme for the I<sup>2</sup>C interface combines the slave device address and device register addresses into a single 7-bit address as shown below.

$$\text{ADDRESS}[6:0] = (\text{I2C\_GROUP\_ADDR}[3:0] \ll 3) + \text{REG\_ADDR}[4:0]$$

The I2C\_GROUP\_ADDR is a 4-bit value stored in the EEPROM. REG\_ADDR is the 5-bit register address being accessed, and can range from 0x00 – 0x1F. The factory programmed value of the group address is '0100'. Contact TI if an alternative group address is required.

For the default I2C\_GROUP\_ADDR, the combined address can be formed as shown in [Table 4](#).

**Table 4. Combined I<sup>2</sup>C Address for Default Group Address**

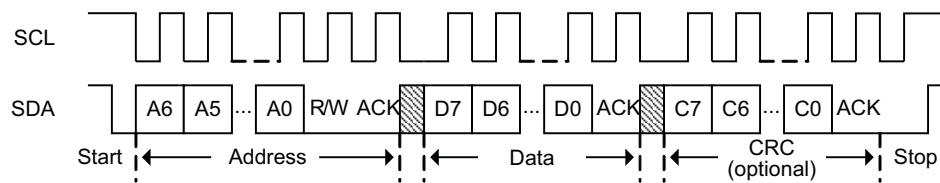
| ADDRESS[6:0] |   |                  |
|--------------|---|------------------|
| 6            | 5 | 4:0              |
| 0            | 1 | Register address |

### 8.5.1.2 Bus Write Command to bq76925

The Host writes to the registers of the bq76925 device as shown in Figure 13. The bq76925 acknowledges each received byte by pulling the SDA line low during the acknowledge period.

The Host may optionally send a CRC after the Data byte as shown. The CRC for write commands is enabled by writing the CRC\_EN bit in the CONFIG\_2 register. If the CRC is not used, then the Host generates the Stop condition immediately after the bq76925 acknowledges receipt of the Data byte.

When the CRC is disabled, the bq76925 device will act on the command on the first rising edge of SCL following the ACK of the Data byte. This occurs as part of the normal bus setup prior to a Stop. If a CRC byte is sent while the CRC is disabled, the first rising edge of the SCL following the ACK will be the clocking of the first bit of the CRC. The bq76925 device does not distinguish these two cases. In both cases, the command will complete normally, and in the latter case the CRC will be ignored.

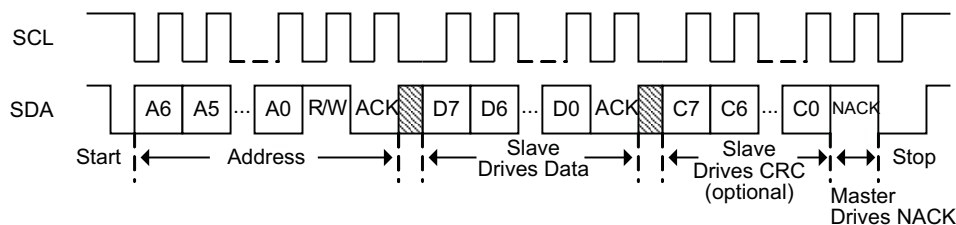


**Figure 13. I²C Write Command**

### 8.5.1.3 Bus Read Command from bq76925 Device

The Host reads from the registers of the bq76925 device as shown in Figure 14. This protocol is similar to the write protocol, except that the slave now drives data back to the Host. The bq76925 device acknowledges each received byte by pulling the SDA line low during the acknowledge period. When the bq76925 device sends data back to the Host, the Host drives the acknowledge.

The Host may optionally request a CRC byte following the Data byte as shown. The CRC for read commands is always enabled, but not required. If the CRC is not used, then the Host simply NACK's the Data byte and then generates the Stop condition.



**Figure 14. I²C Read Command**

## 8.6 Register Maps

| Address     | Name         | Access | D7               | D6        | D5        | D4        | D3             | D2        | D1        | D0        |
|-------------|--------------|--------|------------------|-----------|-----------|-----------|----------------|-----------|-----------|-----------|
| 0x00        | STATUS       | R/W    |                  |           |           |           |                | ALERT     | CRC_ERR   | POR       |
| 0x01        | CELL_CTL     | R/W    |                  |           | VCOUT_SEL |           |                | CELL_SEL  |           |           |
| 0x02        | BAL_CTL      | R/W    |                  |           | BAL_6     | BAL_5     | BAL_4          | BAL_3     | BAL_2     | BAL_1     |
| 0x03        | CONFIG_1     | R/W    | I_THRESH         |           |           |           | I_COMP_POL     | I_AMP_CAL |           | I_GAIN    |
| 0x04        | CONFIG_2     | R/W    | CRC_EN           |           |           |           |                |           |           | REF_SEL   |
| 0x05        | POWER_CTL    | R/W    | SLEEP            | SLEEP_DIS |           | I_COMP_EN | I_AMP_EN       | VC_AMP_EN | VTB_EN    | REF_EN    |
| 0x06        | Reserved     | R/W    |                  |           |           |           |                |           |           |           |
| 0x07        | CHIP_ID      | RO     | CHIP_ID          |           |           |           |                |           |           |           |
| 0x08 – 0x0F | Reserved     | R/W    |                  |           |           |           |                |           |           |           |
| 0x10        | VREF_CAL     | EEPROM | VREF_OFFSET_CORR |           |           |           | VREF_GAIN_CORR |           |           |           |
| 0x11        | VC1_CAL      | EEPROM | VC1_OFFSET_CORR  |           |           |           | VC1_GAIN_CORR  |           |           |           |
| 0x12        | VC2_CAL      | EEPROM | VC2_OFFSET_CORR  |           |           |           | VC2_GAIN_CORR  |           |           |           |
| 0x13        | VC3_CAL      | EEPROM | VC3_OFFSET_CORR  |           |           |           | VC3_GAIN_CORR  |           |           |           |
| 0x14        | VC4_CAL      | EEPROM | VC4_OFFSET_CORR  |           |           |           | VC4_GAIN_CORR  |           |           |           |
| 0x15        | VC5_CAL      | EEPROM | VC5_OFFSET_CORR  |           |           |           | VC5_GAIN_CORR  |           |           |           |
| 0x16        | VC6_CAL      | EEPROM | VC6_OFFSET_CORR  |           |           |           | VC6_GAIN_CORR  |           |           |           |
| 0x17        | VC_CAL_EXT_1 | EEPROM | VC1_OC_4         | VC1_GC_4  | VC2_OC_4  | VC2_GC_4  |                |           |           |           |
| 0x18        | VC_CAL_EXT_2 | EEPROM | VC3_OC_4         | VC3_GC_4  | VC4_OC_4  | VC4_GC_4  | VC5_OC_4       | VC5_GC_4  | VC6_OC_4  | VC6_GC_4  |
| 0x10 – 0x1A | Reserved     | EEPROM |                  |           |           |           |                |           |           |           |
| 0x1B        | VREF_CAL_EXT | EEPROM |                  |           |           |           | 1              | VREF_OC_5 | VREF_OC_4 | VREF_GC_4 |
| 0x1C – 0x1F | Reserved     | EEPROM |                  |           |           |           |                |           |           |           |

### 8.6.1 Register Descriptions

**Table 5. STATUS Register**

| Address   | Name   | Type | D7 | D6 | D5 | D4 | D3 | D2    | D1      | D0  |
|-----------|--------|------|----|----|----|----|----|-------|---------|-----|
| 0x00      | STATUS | R/W  |    |    |    |    |    | ALERT | CRC_ERR | POR |
| Defaults: |        |      | 0  | 0  | 0  | 0  | 0  | 0     | 0       | 1   |

ALERT: Over-current alert. Reflects state of the over-current comparator. '1' = over-current.

CRC\_ERR: CRC error status. Updated on every I<sup>2</sup>C write packet when CRC\_EN = '1'. '1' = CRC error.

POR: Power on reset flag. Set on each power-up and wake-up from sleep. May be cleared by writing with '0'.

**Table 6. CELL\_CTL**

| Address   | Name     | Type | D7 <sup>(1)</sup> | D6 | D5        | D4 | D3 | D2       | D1 | D0 |
|-----------|----------|------|-------------------|----|-----------|----|----|----------|----|----|
| 0x01      | CELL_CTL | R/W  |                   |    | VCOUT_SEL |    |    | CELL_SEL |    |    |
| Defaults: |          |      | 0                 | 0  | 0         | 0  | 0  | 0        | 0  | 0  |

(1) This bit must be kept = 0

VCOUT\_SEL: VCOUT MUX select. Selects the VCOUT pin function as follows.

**Table 7. VCOUT Pin Functions**

| VCOUT_SEL | VCOUT                          |
|-----------|--------------------------------|
| 0 0       | VSS                            |
| 0 1       | VCn (n determined by CELL_SEL) |
| 1 0       | VREF × 0.5                     |
| 1 1       | VREF × 0.85                    |

CELL\_SEL: Cell select. Selects the VCn input for output on VCOUT when VCOUT\_SEL = '01'.

**Table 8. Cell Selection**

| VCOUT_SEL | CELL_SEL | VCOUT                 |
|-----------|----------|-----------------------|
| 0 1       | 0 0 0    | VC1                   |
| 0 1       | 0 0 1    | VC2                   |
| 0 1       | 0 1 0    | VC3                   |
| 0 1       | 0 1 1    | VC4                   |
| 0 1       | 1 0 0    | VC5                   |
| 0 1       | 1 0 1    | VC6                   |
| 0 1       | 1 1 0    | V <sub>TEMP,INT</sub> |
| 0 1       | 1 1 1    | Hi-Z                  |

**Table 9. BAL\_CTL**

| Address   | Name    | Type | D7 | D6 | D5    | D4    | D3    | D2    | D1    | D0    |
|-----------|---------|------|----|----|-------|-------|-------|-------|-------|-------|
| 0x02      | BAL_CTL | R/W  |    |    | BAL_6 | BAL_5 | BAL_4 | BAL_3 | BAL_2 | BAL_1 |
| Defaults: |         |      | 0  | 0  | 0     | 0     | 0     | 0     | 0     | 0     |

BAL\_n: Balance control for cell n. When set, turns on balancing transistor for cell n. Setting of two adjacent balance controls is not permitted. If two adjacent balance controls are set, neither cell balancing transistor will be turned on. However, the BAL\_n bits will retain their values.

**Table 10. CONFIG\_1**

| Address   | Name     | Type | D7       | D6 | D5 | D4         | D3        | D2 | D1 | D0     |
|-----------|----------|------|----------|----|----|------------|-----------|----|----|--------|
| 0x03      | CONFIG_1 | R/W  | I_THRESH |    |    | I_COMP_POL | I_AMP_CAL |    |    | I_GAIN |
| Defaults: |          |      | 0        |    |    | 0          | 0         | 0  | 0  | 0      |

I\_THRESH: Current comparator threshold. Sets the threshold of the current comparator as follows:

**Table 11. Current Comparator Threshold**

| I_THRESH | Comparator Threshold |
|----------|----------------------|
| 0x0      | 25 mV                |
| 0x1      | 50 mV                |
| 0x2      | 75 mV                |
| 0x3      | 100 mV               |
| 0x4      | 125 mV               |
| 0x5      | 150 mV               |
| 0x6      | 175 mV               |
| 0x7      | 200 mV               |
| 0x8      | 225 mV               |
| 0x9      | 250 mV               |
| 0xA      | 275 mV               |
| 0xB      | 300 mV               |
| 0xC      | 325 mV               |
| 0xD      | 350 mV               |
| 0xE      | 375 mV               |
| 0xF      | 400 mV               |

I\_COMP\_POL: Current comparator polarity select. When '0', trips on discharge current (SENSEP > SENSEN). When '1', trips on charge current (SENSEP < SENSEN).

I\_AMP\_CAL: Current amplifier calibration. When '0', current amplifier reports SENSEN with respect to VSS. When '1', current amplifier reports SENSEP with respect to VSS. This bit can be used for offset cancellation as described under OPERATIONAL OVERVIEW.

I\_GAIN: Current amplifier gain. Sets the nominal gain of the current amplifier as follows.

**Table 12. Nominal Gain of the Current Amplifier**

| I_GAIN | Current amp gain |
|--------|------------------|
| 0      | 4                |
| 1      | 8                |

**Table 13. CONFIG\_2**

| Address   | Name     | Type | D7     | D6 | D5 | D4 | D3 | D2 | D1 | D0      |
|-----------|----------|------|--------|----|----|----|----|----|----|---------|
| 0x04      | CONFIG_2 | R/W  | CRC_EN |    |    |    |    |    |    | REF_SEL |
| Defaults: |          |      | 0      | 0  | 0  | 0  | 0  | 0  | 0  | 0       |

CRC\_EN: CRC enable. Enables CRC comparison on write. When '1', CRC is enabled. CRC on read is always enabled but is optional for Host.

REF\_SEL: Reference voltage selection. Sets reference voltage output on VREF pin, cell-voltage amplifier gain and VIOUT output range.

**Table 14. Reference Voltage Selection**

| REF_SEL | VREF (V) | VCOUT Gain | VIOUT Output Range (V) |
|---------|----------|------------|------------------------|
| 0       | 1.5      | 0.3        | 0.25 – 1.25            |
| 1       | 3.0      | 0.6        | 0.5 – 2.5              |

**Table 15. POWER\_CTL**

| Address   | Name      | Type | D7    | D6        | D5 | D4        | D3       | D2        | D1     | D0     |
|-----------|-----------|------|-------|-----------|----|-----------|----------|-----------|--------|--------|
| 0x05      | POWER_CTL | R/W  | SLEEP | SLEEP_DIS |    | I_COMP_EN | I_AMP_EN | VC_AMP_EN | VTB_EN | REF_EN |
| Defaults: |           |      | 0     | 0         | 0  | 0         | 0        | 0         | 0      | 0      |

SLEEP: Sleep control. Set to '1' to put device to sleep

SLEEP\_DIS: Sleep mode disable. When '1', disables the sleep mode.

I\_COMP\_EN: Current comparator enable. When '1', comparator is enabled. Disable to save power.

I\_AMP\_EN: Current amplifier enable. When '1', current amplifier is enabled. Disable to save power.

VC\_AMP\_EN: Cell amplifier enable. When '1', cell amplifier is enabled. Disable to save power.

VTB\_EN: Thermistor bias enable. When '1', the VTB pin is internally switched to the V3P3 voltage.

REF\_EN: Voltage reference enable. When '1', the 1.5 / 3.0 V reference is enabled. Disable to save power

**Table 16. CHIP\_ID**

| Address   | Name    | Type | D7      | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|-----------|---------|------|---------|----|----|----|----|----|----|----|
| 0x07      | CHIP_ID | RO   | CHIP_ID |    |    |    |    |    |    |    |
| Defaults: |         |      | 0x10    |    |    |    |    |    |    |    |

CHIP\_ID: Silicon version identifier.

**Table 17. VREF\_CAL**

| Address | Name     | Type   | D7               | D6 | D5 | D4 | D3             | D2 | D1 | D0 |
|---------|----------|--------|------------------|----|----|----|----------------|----|----|----|
| 0x10    | VREF_CAL | EEPROM | VREF_OFFSET_CORR |    |    |    | VREF_GAIN_CORR |    |    |    |

**VREF\_OFFSET\_CORR:** Lower 4 bits of offset-correction factor for reference output. The complete offset-correction factor is obtained by concatenating this value with the the two most significant bits VREF\_OC\_5 and VREF\_OC\_4, which are stored in the VREF\_CAL\_EXT register. The final value is a 6-bit signed 2's complement number in the range –32 to +31 with a value of 1 mV per LSB. See description of usage in [Detailed Description](#).

**VREF\_GAIN\_CORR:** Lower 4 bits of gain correction factor for reference output. The complete gain correction factor is obtained by concatenating this value with the most significant bit VREF\_GC\_4, which is stored in the VREF\_CAL\_EXT register. The final value is a 5-bit signed 2's complement number in the range –16 to +15 with a value of 0.1% per lsb. See description of usage in [Detailed Description](#).

**Table 18. VC1\_CAL**

| Address | Name    | Type   | D7              | D6 | D5 | D4 | D3            | D2 | D1 | D0 |
|---------|---------|--------|-----------------|----|----|----|---------------|----|----|----|
| 0x11    | VC1_CAL | EEPROM | VC1_OFFSET_CORR |    |    |    | VC1_GAIN_CORR |    |    |    |

**VC1\_OFFSET\_CORR:** Lower 4 bits of offset correction factor for cell 1 translation. The complete offset correction factor is obtained by concatenating this value with the most significant bit VC1\_OC\_4, which is stored in the VC\_CAL\_EXT\_1 register. The final value is a 5-bit signed 2's complement number in the range –16 to +15 with a value of 1 mV per lsb. See description of usage in [Detailed Description](#).

**VC1\_GAIN\_CORR:** Lower 4 bits of gain correction factor for cell 1 translation. The complete gain correction factor is obtained by concatenating this value with the most significant bit VC1\_GC\_4, which is stored in the VC\_CAL\_EXT\_1 register. The final value is a 5-bit signed 2's complement number in the range –16 to +15 with a value of 0.1% per lsb. See description of usage in [Detailed Description](#).

**Table 19. VC2\_CAL**

| Address | Name    | Type   | D7              | D6 | D5 | D4 | D3            | D2 | D1 | D0 |
|---------|---------|--------|-----------------|----|----|----|---------------|----|----|----|
| 0x12    | VC2_CAL | EEPROM | VC2_OFFSET_CORR |    |    |    | VC2_GAIN_CORR |    |    |    |

**VC2\_OFFSET\_CORR:** Lower 4 bits of offset correction factor for cell 2 translation. The complete offset correction factor is obtained by concatenating this value with the most significant bit VC2\_OC\_4, which is stored in the VC\_CAL\_EXT\_1 register. The final value is a 5-bit signed 2's complement number in the range –16 to +15 with a value of 1 mV per LSB. See description of usage in [Detailed Description](#).

**VC2\_GAIN\_CORR:** Lower 4 bits of gain correction factor for cell 2 translation. The complete gain correction factor is obtained by concatenating this value with the most significant bit VC2\_GC\_4, which is stored in the VC\_CAL\_EXT\_1 register. The final value is a 5-bit signed 2's complement number in the range –16 to +15 with a value of 0.1% per LSB. See description of usage in [Detailed Description](#).

**Table 20. VC3\_CAL**

| Address | Name    | Type   | D7              | D6 | D5 | D4 | D3            | D2 | D1 | D0 |
|---------|---------|--------|-----------------|----|----|----|---------------|----|----|----|
| 0x13    | VC3_CAL | EEPROM | VC3_OFFSET_CORR |    |    |    | VC3_GAIN_CORR |    |    |    |

**VC3\_OFFSET\_CORR:** Lower 4 bits of offset correction factor for cell 3 translation. The complete offset correction factor is obtained by concatenating this value with the most significant bit VC3\_OC\_4, which is stored in the VC\_CAL\_EXT\_2 register. The final value is a 5-bit signed 2's complement number in the range –16 to +15 with a value of 1 mV per lsb. See description of usage in [Detailed Description](#).

**VC3\_GAIN\_CORR:** Lower 4 bits of gain correction factor for cell 3 translation. The complete gain correction factor is obtained by concatenating this value with the most significant bit VC3\_GC\_4, which is stored in the VC\_CAL\_EXT\_2 register. The final value is a 5-bit signed 2's complement number in the range –16 to +15 with a value of 0.1% per lsb. See description of usage in [Detailed Description](#).

**Table 21. VC4\_CAL**

| Address | Name    | Type   | D7              | D6 | D5 | D4 | D3            | D2 | D1 | D0 |
|---------|---------|--------|-----------------|----|----|----|---------------|----|----|----|
| 0x14    | VC4_CAL | EEPROM | VC4_OFFSET_CORR |    |    |    | VC4_GAIN_CORR |    |    |    |

VC4\_OFFSET\_CORR: Lower 4 bits of offset correction factor for cell 4 translation. The complete offset correction factor is obtained by concatenating this value with the most significant bit VC4\_OC\_4, which is stored in the VC\_CAL\_EXT\_2 register. The final value is a 5-bit signed 2's complement number in the range –16 to +15 with a value of 1 mV per lsb. See description of usage in [Detailed Description](#).

VC4\_GAIN\_CORR: Lower 4 bits of gain correction factor for cell 4 translation. The complete gain correction factor is obtained by concatenating this value with the most significant bit VC4\_GC\_4, which is stored in the VC\_CAL\_EXT\_2 register. The final value is a 5-bit signed 2's complement number in the range –16 to +15 with a value of 0.1% per lsb. See description of usage in [Detailed Description](#).

**Table 22. VC5\_CAL**

| Address | Name    | Type   | D7              | D6 | D5 | D4 | D3            | D2 | D1 | D0 |
|---------|---------|--------|-----------------|----|----|----|---------------|----|----|----|
| 0x15    | VC5_CAL | EEPROM | VC5_OFFSET_CORR |    |    |    | VC5_GAIN_CORR |    |    |    |

VC5\_OFFSET\_CORR: Lower 4 bits of offset correction factor for cell 5 translation. The complete offset correction factor is obtained by concatenating this value with the most significant bit VC5\_OC\_4, which is stored in the VC\_CAL\_EXT\_2 register. The final value is a 5-bit signed 2's complement number in the range –16 to +15 with a value of 1 mV per LSB. See description of usage in [Detailed Description](#).

VC5\_GAIN\_CORR: Lower 4 bits of gain correction factor for cell 5 translation. The complete gain correction factor is obtained by concatenating this value with the most significant bit VC5\_GC\_4, which is stored in the VC\_CAL\_EXT\_2 register. The final value is a 5-bit signed 2's complement number in the range –16 to +15 with a value of 0.1% per LSB. See description of usage in [Detailed Description](#).

**Table 23. VC6\_CAL**

| Address | Name    | Type   | D7              | D6 | D5 | D4 | D3            | D2 | D1 | D0 |
|---------|---------|--------|-----------------|----|----|----|---------------|----|----|----|
| 0x16    | VC6_CAL | EEPROM | VC6_OFFSET_CORR |    |    |    | VC6_GAIN_CORR |    |    |    |

VC6\_OFFSET\_CORR: Lower 4 bits of offset correction factor for cell 6 translation. The complete offset correction factor is obtained by concatenating this value with the most significant bit VC6\_OC\_4, which is stored in the VC\_CAL\_EXT\_2 register. The final value is a 5-bit signed 2's complement number in the range –16 to +15 with a value of 1 mV per LSB. See description of usage in [Detailed Description](#).

VC6\_GAIN\_CORR: Lower 4 bits of gain correction factor for cell 6 translation. The complete gain correction factor is obtained by concatenating this value with the most significant bit VC6\_GC\_4, which is stored in the VC\_CAL\_EXT\_2 register. The final value is a 5-bit signed 2's complement number in the range –16 to +15 with a value of 0.1% per LSB. See description of usage in [Detailed Description](#).

**Table 24. VC\_CAL\_EXT\_1**

| Address | Name         | Type   | D7       | D6       | D5       | D4       | D3 | D2 | D1 | D0 |
|---------|--------------|--------|----------|----------|----------|----------|----|----|----|----|
| 0x17    | VC_CAL_EXT_1 | EEPROM | VC1_OC_4 | VC1_GC_4 | VC2_OC_4 | VC2_GC_4 |    |    |    |    |

VC1\_OC\_4: Most significant bit of offset correction factor for cell 1 translation. See [Table 18](#) register description for details.

VC1\_GC\_4: Most significant bit of gain correction factor for cell 1 translation. See [Table 18](#) register description for details.

VC2\_OC\_4: Most significant bit of offset correction factor for cell 2 translation. See [Table 19](#) register description for details.

VC2\_GC\_4: Most significant bit of gain correction factor for cell 2 translation. See [Table 19](#) register description for details.

**Table 25. VC\_CAL\_EXT\_2**

| Address | Name         | Type   | D7       | D6       | D5       | D4       | D3       | D2       | D1       | D0      |
|---------|--------------|--------|----------|----------|----------|----------|----------|----------|----------|---------|
| 0x18    | VC_CAL_EXT_2 | EEPROM | VC3_OC_4 | VC3_GC_4 | VC4_OC_4 | VC4_GC_4 | VC5_OC_4 | VC5_GC_4 | VC6_OC_4 | VC6_GC4 |

VC3\_OC\_4: Most significant bit of offset correction factor for cell 3 translation. See [Table 20](#) register description for details.

VC3\_GC\_4: Most significant bit of gain correction factor for cell 3 translation. See [Table 20](#) register description for details.

VC4\_OC\_4: Most significant bit of offset correction factor for cell 4 translation. See [Table 21](#) register description for details.

VC4\_GC\_4: Most significant bit of gain correction factor for cell 4 translation. See [Table 21](#) register description for details.

VC5\_OC\_4: Most significant bit of offset correction factor for cell 5 translation. See [Table 22](#) register description for details.

VC5\_GC\_4: Most significant bit of gain correction factor for cell 5 translation. See [Table 22](#) register description for details.

VC6\_OC\_4: Most significant bit of offset correction factor for cell 6 translation. See [Table 23](#) register description for details.

VC6\_GC\_4: Most significant bit of gain correction factor for cell 6 translation. See [Table 23](#) register description for details.

**Table 26. VREF\_CAL\_EXT**

| Address | Name         | Type   | D7 | D6 | D5 | D4 | D3 | D2        | D1         | D0       |
|---------|--------------|--------|----|----|----|----|----|-----------|------------|----------|
| 0x1B    | VREF_CAL_EXT | EEPROM |    |    |    |    | 1  | VREF_OC_5 | VCREF_OC_4 | VREF_GC4 |

VREF\_OC\_5: Most significant bit of offset correction factor for reference output. See [Table 17](#) register description for details.

VREF\_OC\_4: Next most significant bit of offset correction factor for reference output. See [Table 17](#) register description for details.

VREF\_GC\_4: Most significant bit of gain correction factor for reference output. See [Table 17](#) register description for details.

## 9 Application and Implementation

### NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

### 9.1 Application Information

The bq76925 device is a host-controlled analog front end (AFE), providing the individual cell voltages, pack current, and temperature to the host system. The host controller may use this information to complete the pack monitoring, balancing, and protection functions for the 3-series to 6-series cell Li-ion/Li-Polymer battery.

The section below highlights several recommended implementations when using this device. A detailed bq76925 Application report, [SLUA619](#), together with an example implementation report using bq76925 and MSP430G2xx2, [SLUA707](#), are available at [www.ti.com](http://www.ti.com).

#### 9.1.1 Recommended System Implementation

##### 9.1.1.1 Voltage, Current, and Temperature Outputs

The bq76925 device provides voltage, current, and temperature outputs in analog form. A microcontroller (MCU) with an analog-to-digital converter (ADC) is required to complete the measurement system. A minimum of three input-ADC channels of the MCU are required to measure cell voltages, current, and temperature output. The bq76925 device can supply an external reference for the MCU ADC reference. Compare the internal reference voltage specification of the MCU to determine if using the AFE reference would improve the measurement accuracy.

##### 9.1.1.2 Power Management

The bq76925 device can disable various functions for power management. Refer to the POWER\_CTL registers in this document for detailed descriptions. Additionally, the MCU can put the bq76925 device into SHUTDOWN mode by writing to the [SLEEP] bit in the POWER\_CTL register. The wake up circuit does not activate until the V3P3 is completing discharge to 0 V. Once the wake up circuit is activated, pulling the ALERT pin high can wake up the device. This means, once the SLEEP command is sent, the bq76925 device remains in SHUTDOWN mode and cannot wake up if V3P3 is > 0 V.

##### 9.1.1.3 Low Dropout (LDO) Regulator

When the LDO load current is higher than 4 mA, the LDO must be used with an external pass transistor. In this configuration, a high-gain bypass device is recommended. ZXTPT2504DFH and IRLML9303 are example transistors. A Z1 diode is recommended to protect the gate-source or base emitter of the bypass transistor.

Adding the  $R_{V3P3}$  and  $C_{V3P3-2}$  filter helps to isolate the load from the V3P3 transient caused by the load and the transients on BAT.

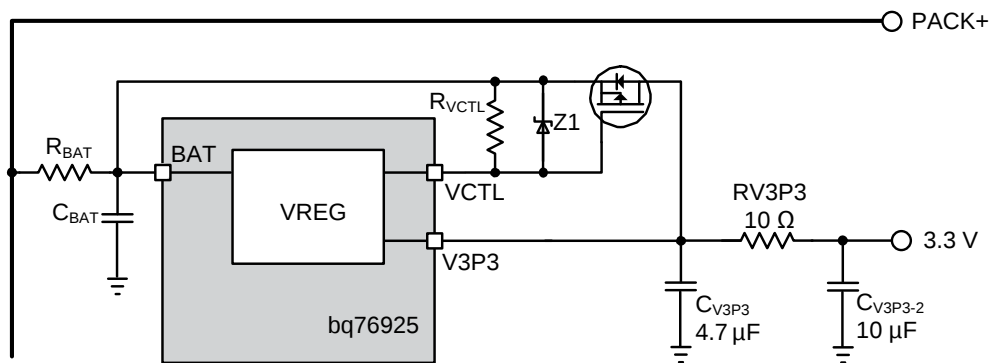


Figure 15. LDO Regulator

## Application Information (continued)

### 9.1.1.4 Input Filters

TI recommends to use input filters for BAT, VCx, and SENSEN/P pins to protect the bq76925 device from large transients caused by switching of the battery load.

Additionally, the filter on BAT also avoids unintentional reset of the AFE when the battery voltage suddenly drops. To further avoid an unwanted reset, a hold-up circuit using a blocking diode can be added in series with the input filter. A zener diode clamp may be added in parallel with the filter capacitor to prevent the repeated peak transients that pump up the filter capacitor beyond the device absolute maximum rating.

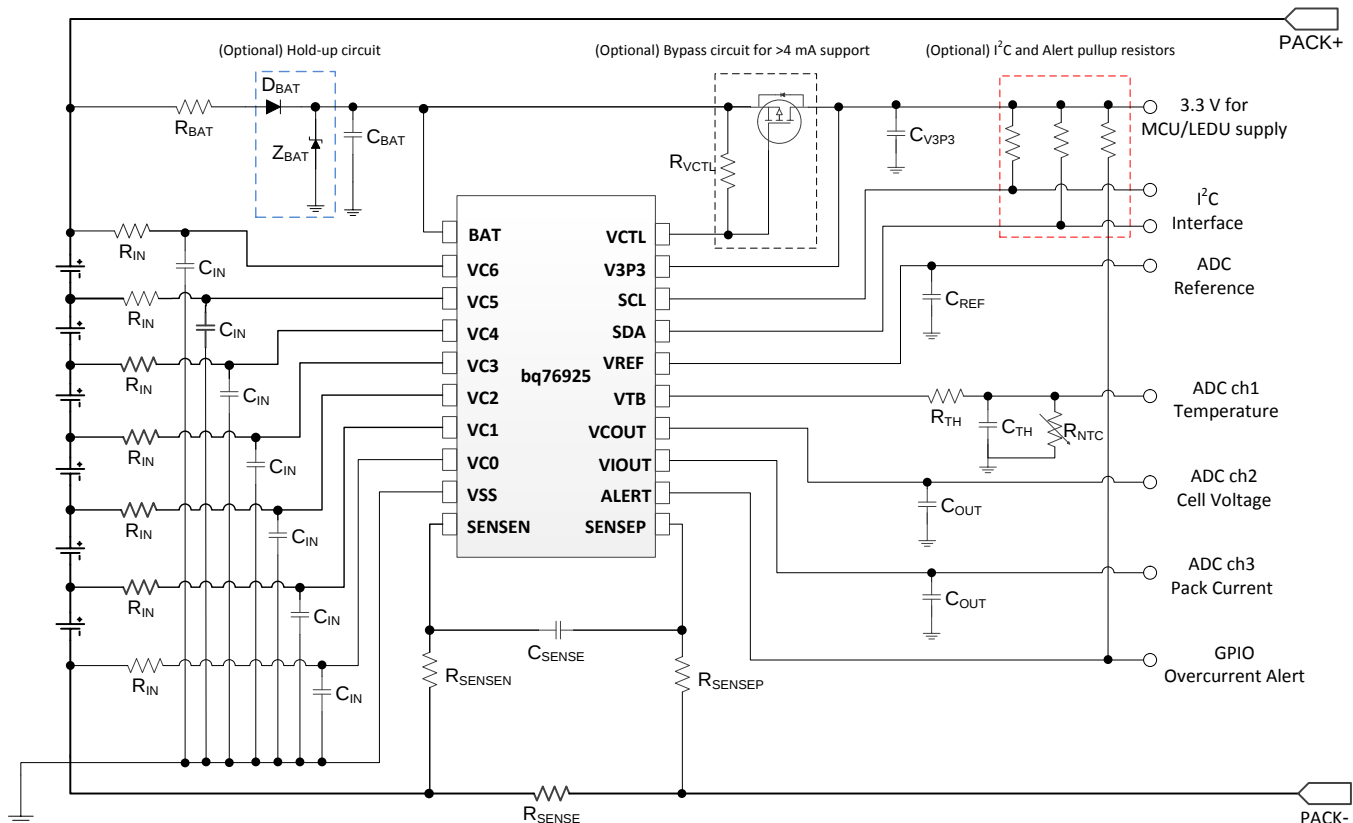
### 9.1.1.5 Output Filters

Output capacitors are used on V3P3, VREF, VCOOUT, and VIOUT for stability. These capacitors also function as bypass capacitors in response to the MCU internal switching and ADC operation. Additional filtering may be added to these output pins to smooth out noisy signals prior to ADC conversion. For the V3P3 case, an additional filter helps reduce the transient on the power input connected to the bq76925 device's V3P3 pin.

### 9.1.2 Cell Balancing

The bq76925 device integrates cell balancing FETs that are controlled individually by the host. The device does not automatically duty cycle the balancing FETs such that cell voltage measurement for protection detection is taken when balancing is off. The host MCU is responsible for such management. Otherwise, the MCU is free to turn on the voltage measurement during cell balancing, which enables the open-cell detection method described in this document. However, the bq76925 device does prevent two adjacent balancing FETs from being turned on simultaneously. If such a condition occurs, both adjacent transistors will remain off.

## 9.2 Typical Application



**Figure 16. Typical Schematic**

## Typical Application (continued)

### 9.2.1 Design Requirements

For this design example, use the parameters listed in [Table 27](#).

**Table 27. Design Parameters**

| PARAMETER                                  |                                        | MIN                                | TYP                | MAX  | UNIT |
|--------------------------------------------|----------------------------------------|------------------------------------|--------------------|------|------|
| R <sub>BAT</sub>                           | BAT filter resistance                  |                                    | 100                |      | Ω    |
| C <sub>BAT</sub>                           | BAT filter capacitance                 |                                    | 10                 |      | μF   |
| R <sub>IN</sub>                            | External cell input resistance         |                                    | <sup>(1)</sup> 100 |      | Ω    |
| C <sub>IN</sub>                            | External cell input capacitance        | 0.1                                | 1                  | 10   | μF   |
| R <sub>SENSEN</sub><br>R <sub>SENSEP</sub> | Current sense input filter resistance  |                                    | 1K                 |      | Ω    |
| C <sub>SENSE</sub>                         | Current sense input filter capacitance |                                    | 0.1                |      | μF   |
| R <sub>VCTL</sub>                          | VCTL pullup resistance                 | Without external bypass transistor |                    | 0    | Ω    |
|                                            |                                        | With external bypass transistor    |                    | 200K |      |
| C <sub>V3P3</sub>                          | V3P3 output capacitance                | Without external bypass transistor |                    | 4.7  | μF   |
|                                            |                                        | With external bypass transistor    |                    | 1.0  |      |
| C <sub>REF</sub>                           | VREF output capacitance                | 1.0                                |                    |      | μF   |
| C <sub>OUT</sub>                           | ADC channel output capacitance         | VCOUT                              |                    | 0.1  | μF   |
|                                            |                                        | VIOUT                              |                    | 470  | 2000 |
|                                            |                                        |                                    |                    |      | pF   |

(1)  $R_{IN,MIN} = 0.5 \times (VCn_{MAX} / 50 \text{ mA})$  if cell balancing used so that maximum recommended cell balancing current is not exceeded.

### 9.2.2 Detailed Design Procedure

The following is the detailed design procedure.

1. Select a proper MCU to complete the battery management solution. Refer to the bq76925 Application report, [SLUA619](#) on MCU requirement.
2. Based on the system design, determine if an alternative cell connection for 4-series and 5- series battery pack is needed. Refer to the “Cell Amplifier Headroom Under BAT Voltage Drop” section of this document.
3. Determine if a hold-up circuit for BAT and/or an external bypass transistor is needed based on the system design. Follow the reference schematic to complete the circuit design.
4. An example circuit design and MCU code implementation is documented in [SLUA707](#) using bq79625 and MSP430G2xx2.

## 9.2.3 Application Curves

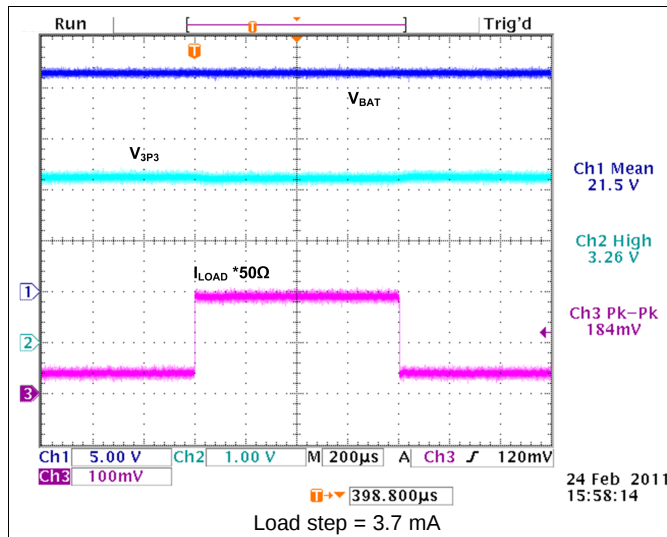


Figure 17. Voltage Regulator With Internal FET

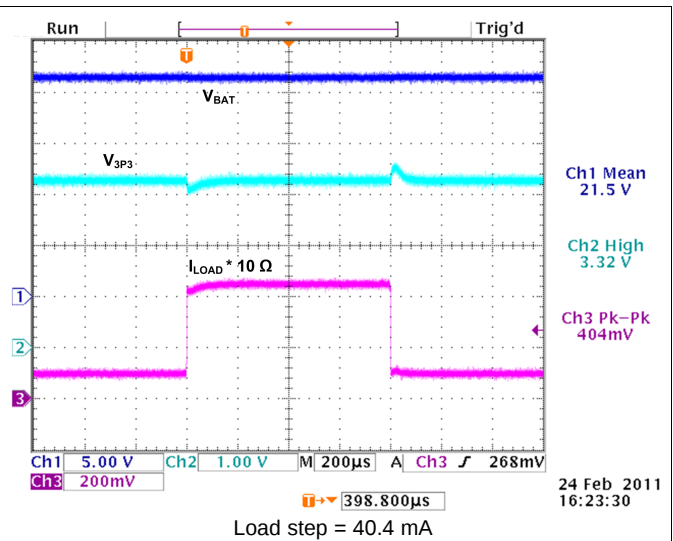


Figure 18. Voltage Regulator With External FET

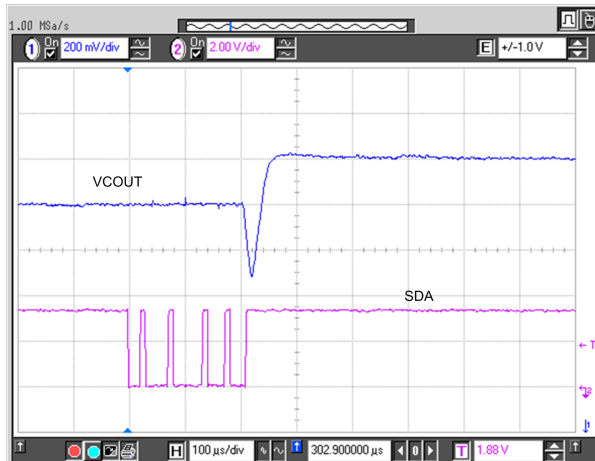


Figure 19. VCOUT Settling With 200 mV Step

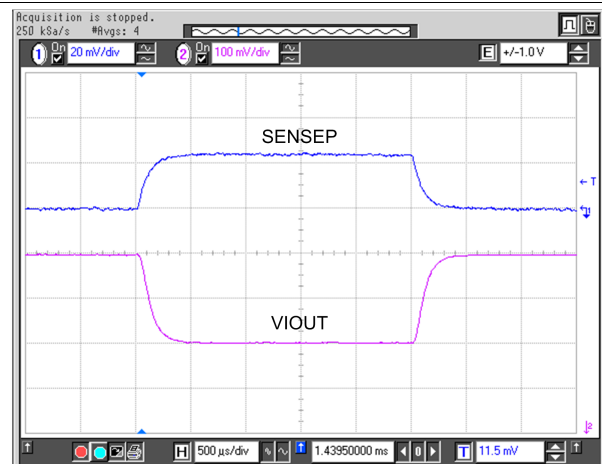


Figure 20. VIOUT Settling With 200 mV Step

## 10 Power Supply Recommendations

The maximum operating voltage on the BAT is 26.4 V. In some cases, a peak transient can be more than twice the battery's DC voltage. Ensure the device does not go beyond its absolute maximum rating.

## 11 Layout

### 11.1 Layout Guidelines

1. Place input filters for BAT, VCx, and SENSEN/P close to the device
2. Place output capacitors on V3P3, VREF, VCOOUT, and VIOOUT close to the device
3. Please output filters (if any) close to the target device (for example, the MCU ADC input ports)
4. Isolate high-current and low-current groundings. The AFE, filter capacitors, and MCU grounds should connect to the low-current ground plane of the PCB.

### 11.2 Layout Example

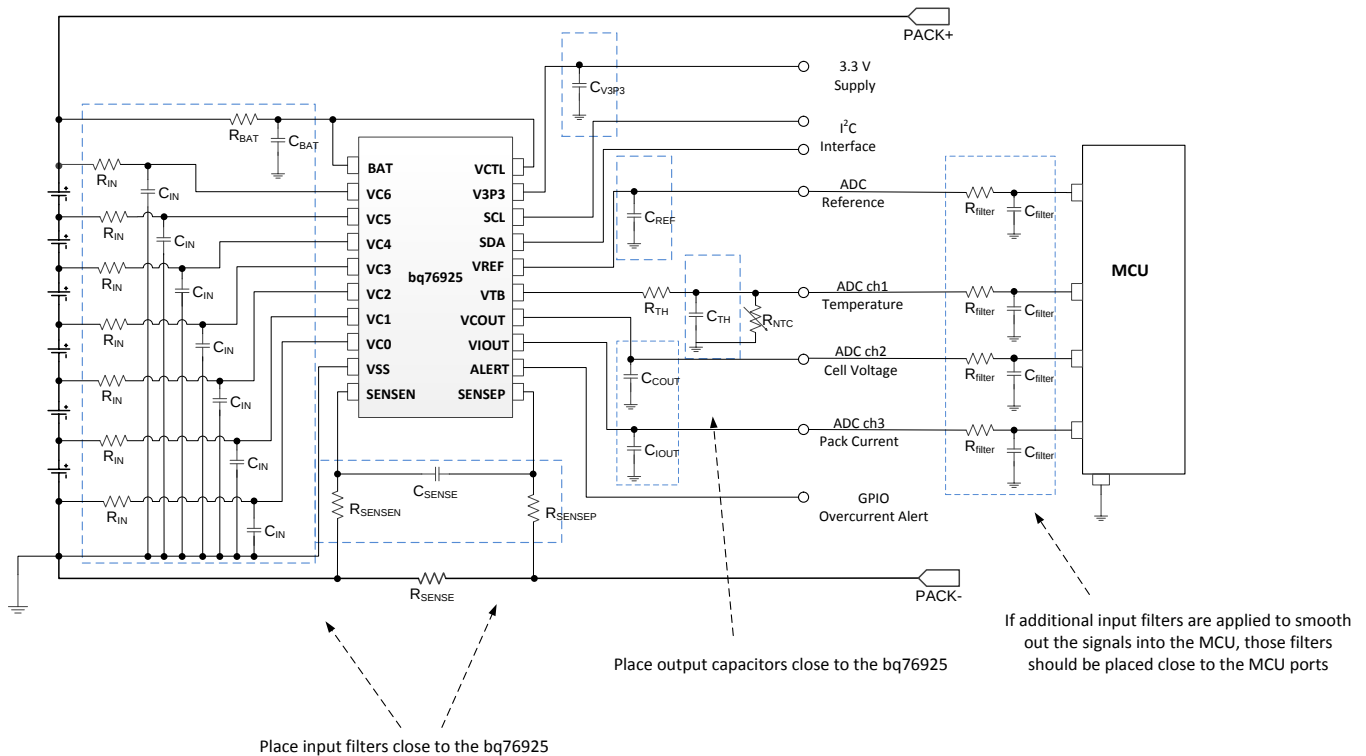


Figure 21. Filters and Bypass Capacitors Placement

## Layout Example (接下页)

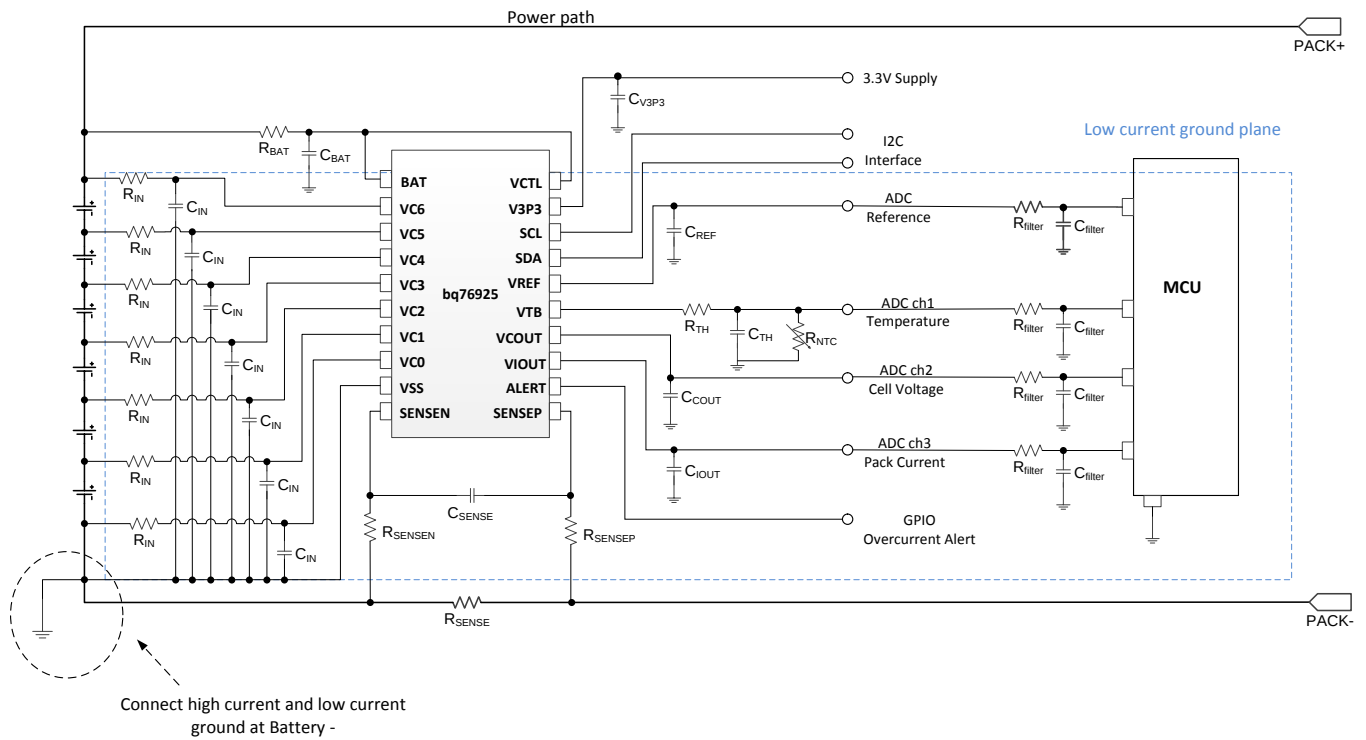


Figure 22. Separate High-Current and Low-Current Grounds

## 12 器件和文档支持

### 12.1 文档支持

#### 12.1.1 相关文档

- 《半导体和 IC 封装热指标》，SPRA953
- 《bq76925 入门》，SLUA619
- 《基于 bq76925 + MSP430G2xx2 的 3 至 6 节电池管理系统》，SLUA707

### 12.2 接收文档更新通知

如需接收文档更新通知，请访问 [www.ti.com.cn](http://www.ti.com.cn) 网站上的器件产品文件夹。点击右上角的提醒我 (Alert me) 注册后，即可每周定期收到已更改的产品信息。有关更改的详细信息，请查阅已修订文档中包含的修订历史记录。

### 12.3 社区资源

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**TI E2E™ Online Community** *TI's Engineer-to-Engineer (E2E) Community*. Created to foster collaboration among engineers. At [e2e.ti.com](http://e2e.ti.com), you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

**Design Support** *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

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这些装置包含有限的内置 ESD 保护。存储或装卸时，应将导线一起截短或将装置放置于导电泡棉中，以防止 MOS 门极遭受静电损伤。

### 12.6 Glossary

**SLYZ022** — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

## 13 机械、封装和可订购信息

以下页中包括机械、封装和可订购信息。这些信息是针对指定器件可提供的最新数据。这些数据会在无通知且不对本文档进行修订的情况下发生改变。欲获得该数据表的浏览器版本，请查阅左侧的导航栏。

## PACKAGING INFORMATION

| Orderable Device | Status<br>(1) | Package Type | Package<br>Drawing | Pins | Package<br>Qty | Eco Plan<br>(2) | Lead finish/<br>Ball material<br>(6) | MSL Peak Temp<br>(3) | Op Temp (°C) | Device Marking<br>(4/5) | Samples                 |
|------------------|---------------|--------------|--------------------|------|----------------|-----------------|--------------------------------------|----------------------|--------------|-------------------------|-------------------------|
| BQ76925PW        | ACTIVE        | TSSOP        | PW                 | 20   | 70             | RoHS & Green    | NIPDAU                               | Level-2-260C-1 YEAR  | -25 to 85    | BQ76925                 | <a href="#">Samples</a> |
| BQ76925PWR       | ACTIVE        | TSSOP        | PW                 | 20   | 2000           | RoHS & Green    | NIPDAU                               | Level-2-260C-1 YEAR  | -25 to 85    | BQ76925                 | <a href="#">Samples</a> |
| BQ76925RGER      | ACTIVE        | VQFN         | RGE                | 24   | 3000           | RoHS & Green    | NIPDAU                               | Level-2-260C-1 YEAR  | -25 to 85    | BQ76925                 | <a href="#">Samples</a> |
| BQ76925RGET      | ACTIVE        | VQFN         | RGE                | 24   | 250            | RoHS & Green    | NIPDAU                               | Level-1-260C-UNLIM   | -25 to 85    | BQ76925                 | <a href="#">Samples</a> |

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "-" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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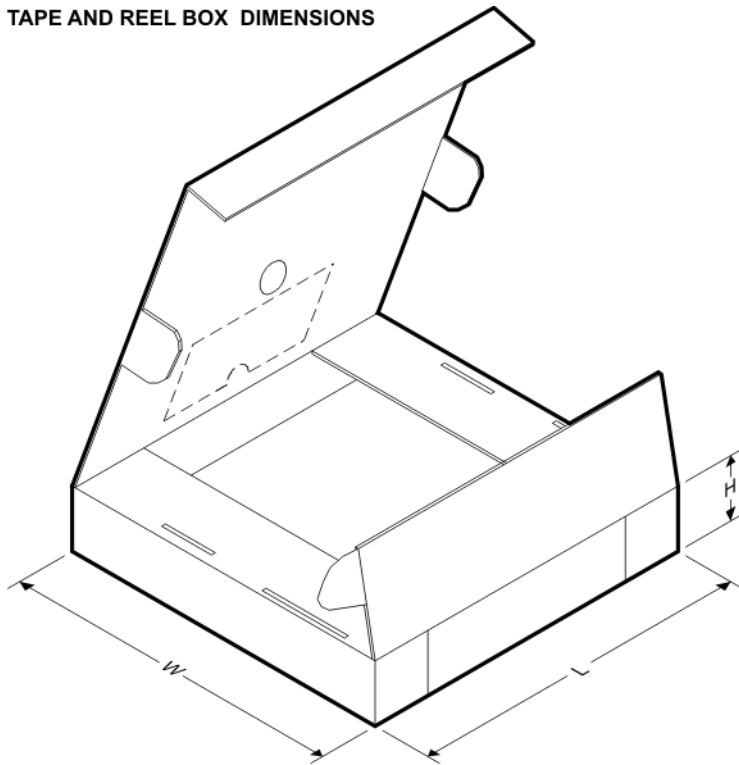
## TAPE AND REEL INFORMATION



\*All dimensions are nominal

| Device      | Package Type | Package Drawing | Pins | SPQ  | Reel Diameter (mm) | Reel Width W1 (mm) | A0 (mm) | B0 (mm) | K0 (mm) | P1 (mm) | W (mm) | Pin1 Quadrant |
|-------------|--------------|-----------------|------|------|--------------------|--------------------|---------|---------|---------|---------|--------|---------------|
| BQ76925PWR  | TSSOP        | PW              | 20   | 2000 | 330.0              | 16.4               | 6.95    | 7.0     | 1.4     | 8.0     | 16.0   | Q1            |
| BQ76925RGER | VQFN         | RGE             | 24   | 3000 | 330.0              | 12.4               | 4.25    | 4.25    | 1.15    | 8.0     | 12.0   | Q2            |
| BQ76925RGET | VQFN         | RGE             | 24   | 250  | 180.0              | 12.4               | 4.25    | 4.25    | 1.15    | 8.0     | 12.0   | Q2            |

## TAPE AND REEL BOX DIMENSIONS



\*All dimensions are nominal

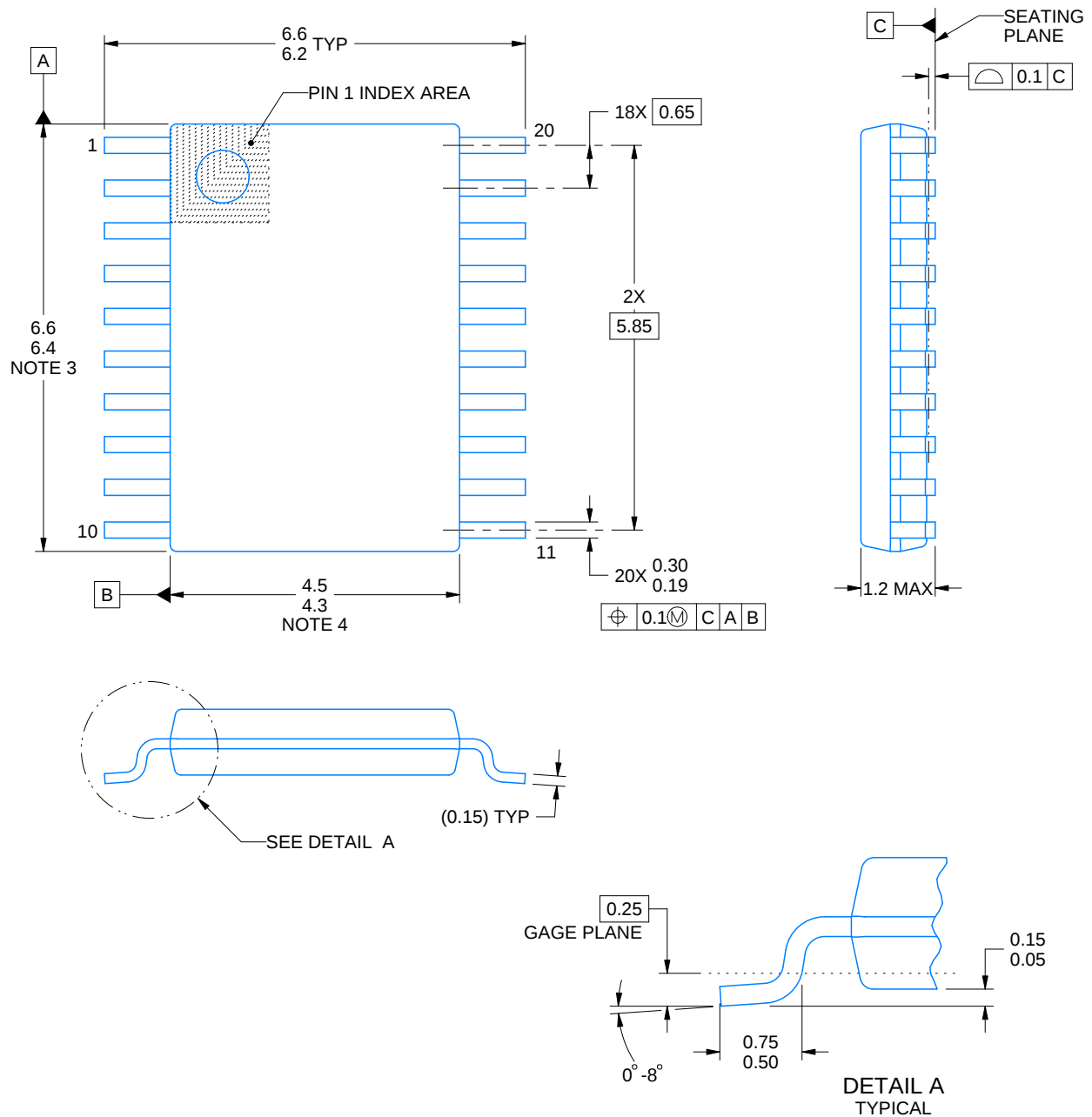
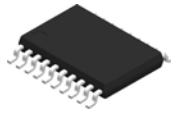
| Device      | Package Type | Package Drawing | Pins | SPQ  | Length (mm) | Width (mm) | Height (mm) |
|-------------|--------------|-----------------|------|------|-------------|------------|-------------|
| BQ76925PWR  | TSSOP        | PW              | 20   | 2000 | 853.0       | 449.0      | 35.0        |
| BQ76925RGER | VQFN         | RGE             | 24   | 3000 | 367.0       | 367.0      | 35.0        |
| BQ76925RGET | VQFN         | RGE             | 24   | 250  | 210.0       | 185.0      | 35.0        |

## TUBE



\*All dimensions are nominal

| Device    | Package Name | Package Type | Pins | SPQ | L (mm) | W (mm) | T (μm) | B (mm) |
|-----------|--------------|--------------|------|-----|--------|--------|--------|--------|
| BQ76925PW | PW           | TSSOP        | 20   | 70  | 530    | 10.2   | 3600   | 3.5    |
| BQ76925PW | PW           | TSSOP        | 20   | 70  | 530    | 10.2   | 3600   | 3.5    |



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## NOTES:

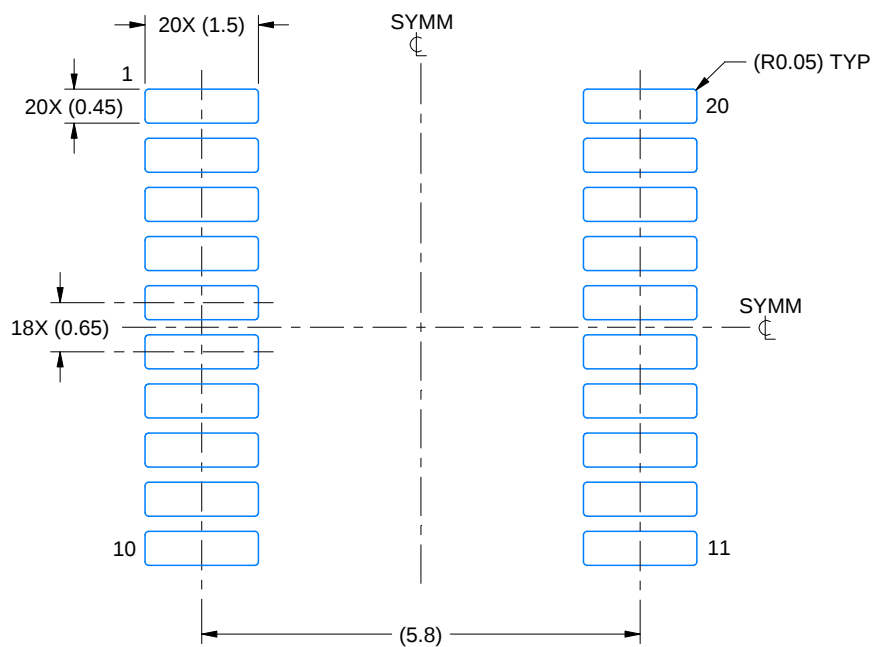
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
5. Reference JEDEC registration MO-153.

# EXAMPLE BOARD LAYOUT

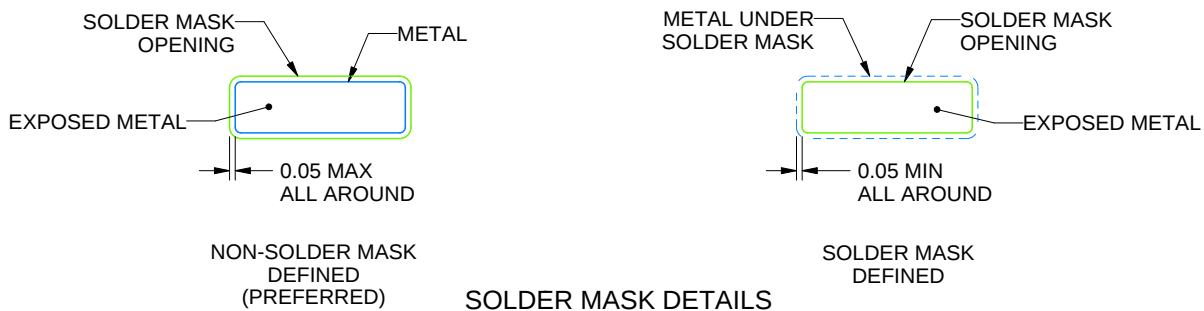
PW0020A

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE: 10X



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NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

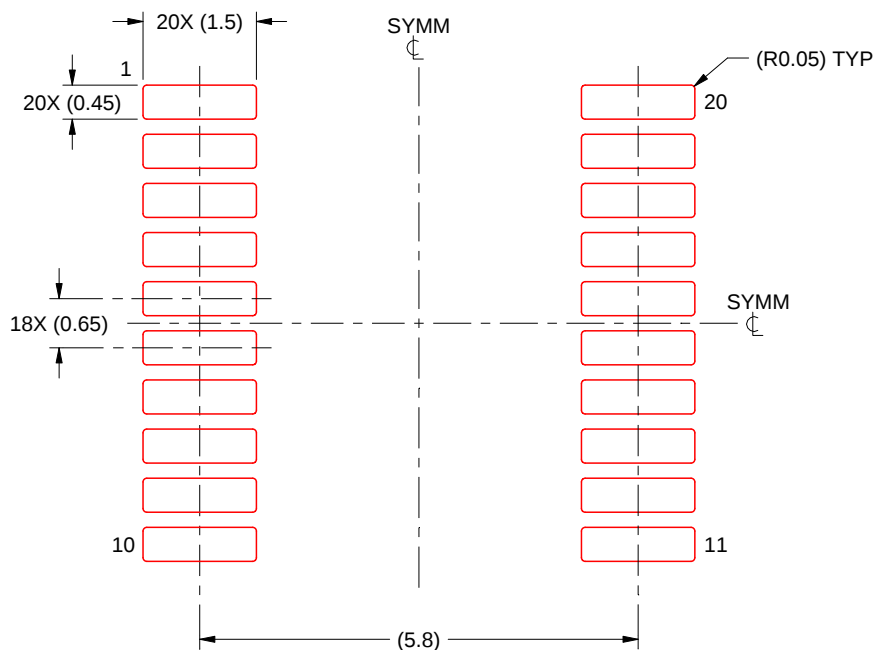
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

## EXAMPLE STENCIL DESIGN

PW0020A

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL  
SCALE: 10X

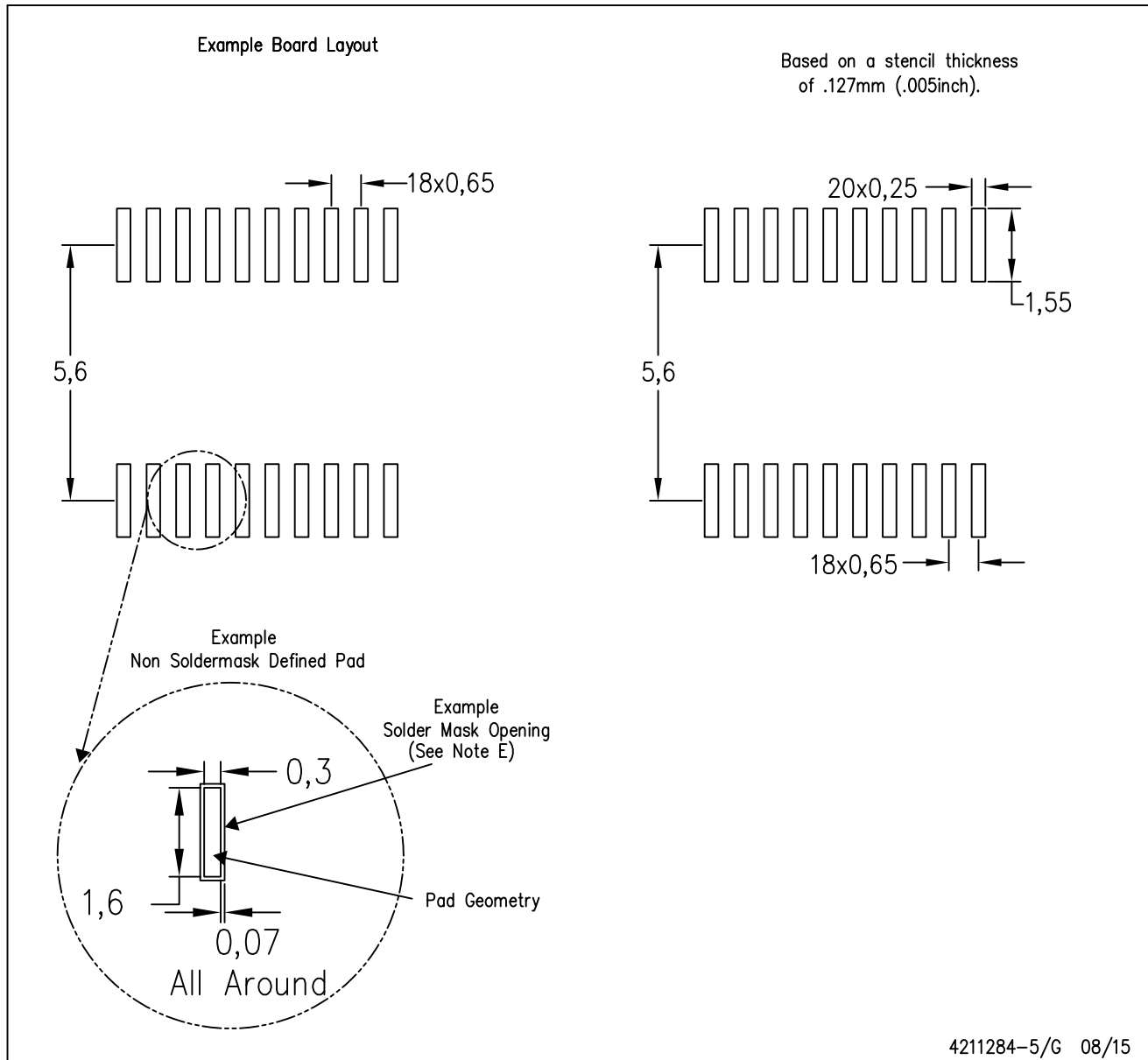
4220206/A 02/2017

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

PW (R-PDSO-G20)

PLASTIC SMALL OUTLINE



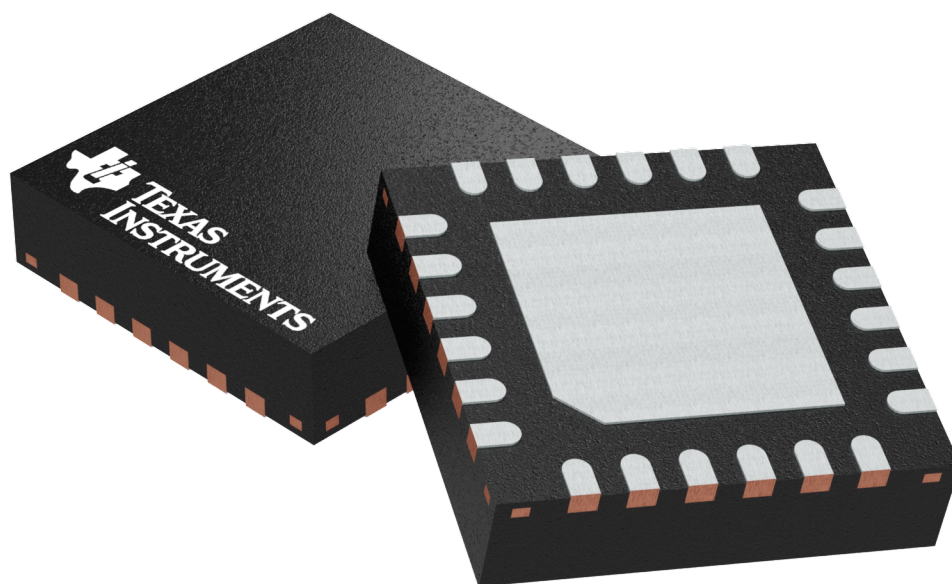
- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Publication IPC-7351 is recommended for alternate design.
  - D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
  - E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

**RGE 24**

**GENERIC PACKAGE VIEW**

**VQFN - 1 mm max height**

PLASTIC QUAD FLATPACK - NO LEAD



Images above are just a representation of the package family, actual package may vary.  
Refer to the product data sheet for package details.

4204104/H



### VQFN - 1 mm max height

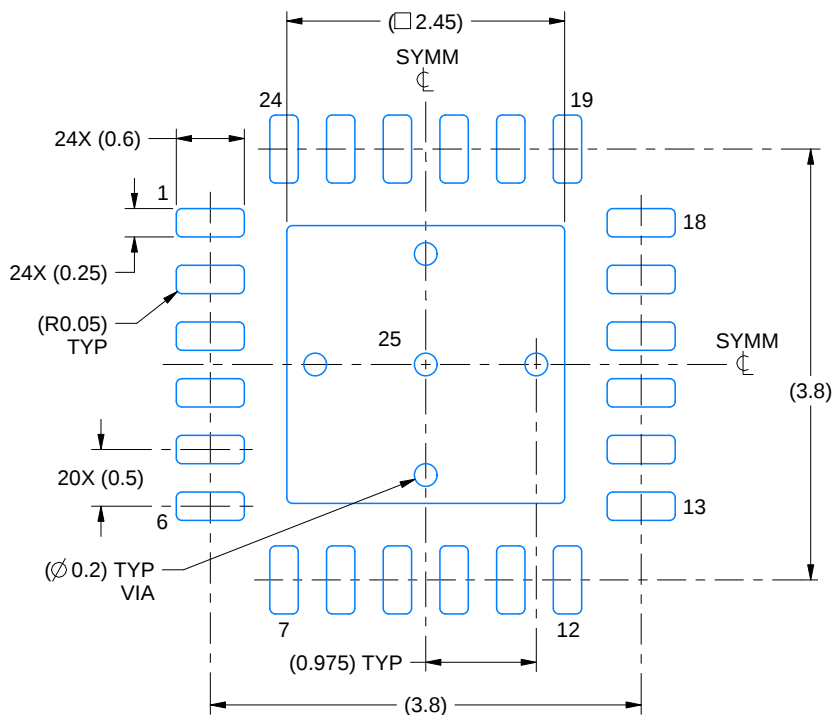
[illegible]

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

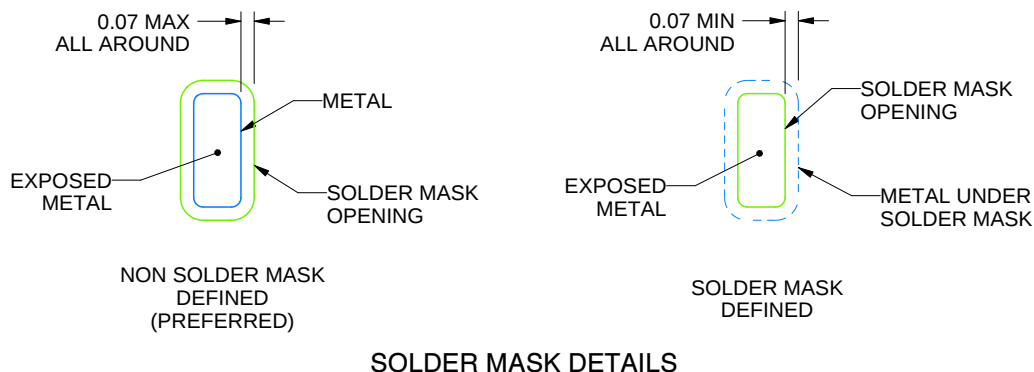
**RGE0024B**

### VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE:15X



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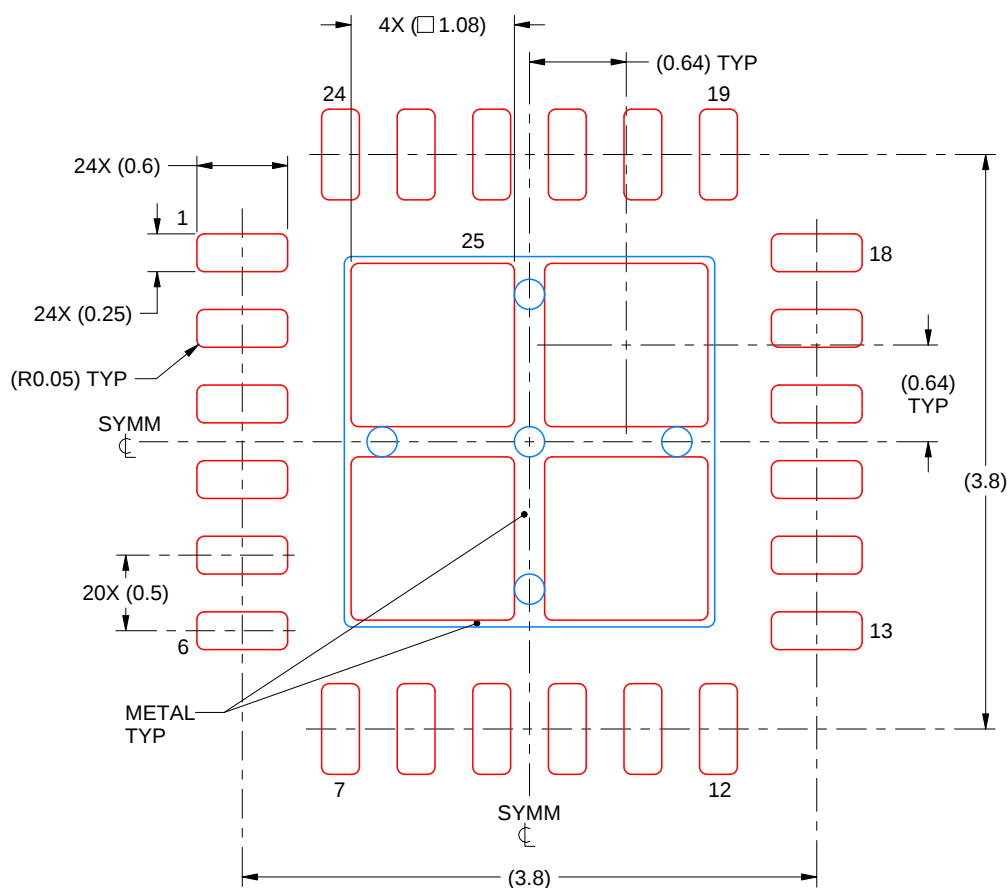
NOTES: (continued)

4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 ([www.ti.com/lit/slua271](http://www.ti.com/lit/slua271)).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

**RGE0024B**

**VQFN - 1 mm max height**

PLASTIC QUAD FLATPACK - NO LEAD



## SOLDER PASTE EXAMPLE BASED ON 0.125 mm THICK STENCIL

EXPOSED PAD 25  
78% PRINTED SOLDER COVERAGE BY AREA UNDER PACKAGE  
SCALE:20X

4219013/A 05/2017

NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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