

UCC27324-Q1 双通道 4A 峰值高速低侧电源 MOSFET 驱动器

1 特性

- 符合汽车类应用的应用
- 行业标准引脚
- 可在米勒平坦区提供 $\pm 4\text{A}$ 的高电流驱动
- 即使在低电源电压下也能实现高效的恒流源
- 与电源电压无关的 TTL 和 CMOS 兼容输入
- 1.8nF 负载时的上升时间和下降时间典型值分别为 20ns 和 15ns
- 输入下降和上升时的典型传播延迟时间分别为 25ns 和 35ns
- 电源电压为 4V 至 15V
- 电源电流为 0.3mA
- 可以并联双输出以获得更高的驱动电流
- T_J 额定值范围为 -40°C 至 125°C
- TrueDrive™ 并联双极和 CMOS 晶体管的输出架构

2 应用

- 开关电源供电
- 直流/直流转换器
- 电机控制器
- 线路驱动器
- D 类开关放大器

3 说明

UCC27324-Q1 高速双路 MOSFET 驱动器可向容性负载提供大峰值电流。通过使用本身能够最大限度减少击穿电流的设计，这些驱动器可在 MOSFET 开关切换期间，在米勒平坦区域提供最需要的 4A 电流。独特的双极和 MOSFET 混合输出级并联，可在低电源电压下实现高效的拉电流和灌电流。

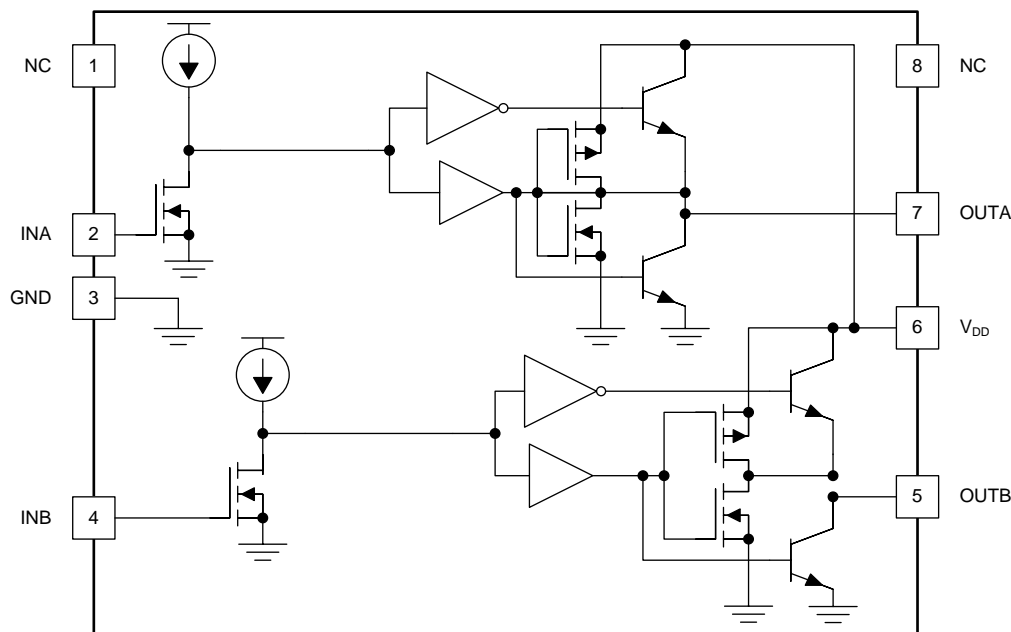
该器件采用标准的 SOIC-8 (D) 封装。

器件信息⁽¹⁾

器件编号	封装	封装尺寸 (标称值)
UCC27324-Q1	SOIC (8)	4.90mm × 3.91mm

(1) 如需了解所有可用封装，请参阅数据表末尾的可订购产品附录。

方框图



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4 修订历史记录

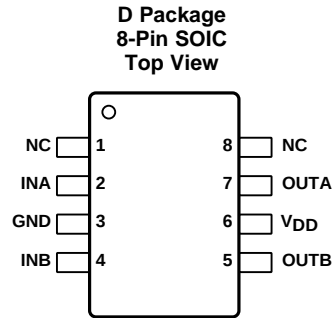
注：之前版本的页码可能与当前版本有所不同。

Changes from Revision B (September 2015) to Revision C	Page
• Changed Figures 8,10,11 to show non-inverting device internally.	12
• Added to equation 6 to correctly multiply by V.	14

Changes from Revision A (April 2012) to Revision B	Page
• 已添加 添加了 ESD 额定值 表、特性 说明 部分、器件功能模式、应用和实施 部分、电源建议 部分、布局部分、器件和文档支持 部分以及机械、封装和可订购信息 部分	1
• 已删除 数据表中的 UCC27323-Q1（双路反相）和 UCC27325-Q1（一路反相、一路同相）器件信息	1
• 已删除 删除了 UCC27324-Q1 器件的 MSOP-PowerPAD 和 PDIP 封装引用	1
• 已删除 删除了 订购信息 表	1
• Deleted <i>Dissipation Ratings</i> table	4

Changes from Original (March, 2008) to Revision A	Page
• Added $T_A = T_J$ to header of Overall Electrical Characteristics table.	4
• Added an extra paragraph before Figure 5.	14

5 Pin Configuration and Functions



NC – No internal connection

Pin Functions

PIN		I/O	DESCRIPTION
NO.	NAME		
1	NC	—	No connection. Should be grounded.
2	INA	I	Input A. Input signal of the A driver. Has logic-compatible threshold and hysteresis. If not used, this input should be tied to either V_{DD} or GND. It should not be left floating.
3	GND	—	Common ground. Should be connected very closely to the source of the power MOSFET that the driver is driving.
4	INB	I	Input B. Input signal of the B driver. Has logic-compatible threshold and hysteresis. If not used, this input should be tied to either V_{DD} or GND. It should not be left floating.
5	OUTB	O	Driver output B. The output stage can provide 4-A drive current to the gate of a power MOSFET.
6	V_{DD}	I	Supply. Supply voltage and the power input connection for this device.
7	OUTA	O	Driver output A. The output stage can provide 4-A drive current to the gate of a power MOSFET.
8	NC	—	No connection. Should be grounded.

6 Specifications

6.1 Absolute Maximum Ratings

 over operating free-air temperature range (unless otherwise noted)⁽¹⁾⁽²⁾

		MIN	MAX	UNIT
V _{DD}	Supply voltage	−0.3	16	V
I _O	Output current (OUTA, OUTB)	DC, I _{OUT_DC}	0.3	A
		Pulsed (0.5 μs), I _{OUT_PULSED}	4.5	
T _J	Junction operating temperature	−55	150	°C
T _{lead}	Lead temperature		300	°C
T _{stg}	Storage temperature, soldering, 10 s	−65	150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltages are with respect to GND. Currents are positive into and negative out of the specified terminal.

6.2 ESD Ratings

		VALUE	UNIT
V _(ESD)	Electrostatic discharge	Human-body model (HBM), per AEC Q100-002 ⁽¹⁾	±2000
		Charged-device model (CDM), per AEC Q100-011	±1000

- (1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

	MIN	NOM	MAX	UNIT
V _{DD}	4		15	V

6.4 Thermal Information

THERMAL METRIC ⁽¹⁾		UCC27324-Q1	UNIT
		D (SOIC)	
		8 PINS	
R _{θJA}	Junction-to-ambient thermal resistance	113	°C/W
R _{θJC(top)}	Junction-to-case (top) thermal resistance	61.7	°C/W
R _{θJB}	Junction-to-board thermal resistance	53.2	°C/W
ψ _{JT}	Junction-to-top characterization parameter	16	°C/W
ψ _{JB}	Junction-to-board characterization parameter	52.7	°C/W
R _{θJC(bot)}	Junction-to-case (bottom) thermal resistance	—	°C/W

- (1) For more information about traditional and new thermal metrics, see the *Semiconductor and IC Package Thermal Metrics* application report, [SPRA953](#).

6.5 Overall Electrical Characteristics

 V_{CC} = 4.5 V to 15 V, T_A = −40°C to 125°C (unless otherwise noted), T_A = T_J

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
I _{DD}	INA = 0 V	INB = 0 V		2	80	μA
		INB = HIGH		300	450	
	INA = HIGH	INB = 0 V		300	450	
		INB = HIGH		600	800	

6.6 Power Dissipation Characteristics

over operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
P_D Power dissipation	$T_A = 25^\circ\text{C}$		650	mW	μA

6.7 Input (INA, INB) Electrical Characteristics

$V_{CC} = 4.5\text{ V to }15\text{ V}$, $T_A = -40^\circ\text{C to }125^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{IH} Logic 1 input threshold		2			V
V_{IL} Logic 0 input threshold				1	V
Input current	$0\text{ V} \leq V_{IN} \leq V_{DD}$	-10	0	10	μA

6.8 Output (OUTA, OUTB) Electrical Characteristics

$V_{CC} = 4.5\text{ V to }15\text{ V}$, $T_A = -40^\circ\text{C to }125^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A	MIN	TYP	MAX	UNIT
Output current ^{(1) (2)}	$V_{DD} = 14\text{ V}$			4		A
V_{OH} High-level output voltage	$V_{OH} = V_{DD} - V_{OUT}$, $I_{OUT} = -10\text{ mA}$			300	450	mV
V_{OL} Low-level output level	$I_{OUT} = 10\text{ mA}$			22	45	mV
Output resistance high ⁽³⁾	$I_{OUT} = -10\text{ mA}$, $V_{DD} = 14\text{ V}$	25°C	25	30	35	Ω
		Full range	18		43	
Output resistance low ⁽³⁾	$I_{OUT} = 10\text{ mA}$, $V_{DD} = 14\text{ V}$	25°C	1.9	2.2	2.5	Ω
		Full range	0.9		4	
Latch-up protection ⁽¹⁾			500			mA

(1) Specified by design

(2) The pullup/pulldown circuits of the driver are bipolar and MOSFET transistors in parallel. The pulsed output current rating is the combined current from the bipolar and MOSFET transistors.

(3) The pullup/pulldown circuits of the driver are bipolar and MOSFET transistors in parallel. The output resistance is the $R_{DS(ON)}$ of the MOSFET transistor when the voltage on the driver output is less than the saturation voltage of the bipolar transistor.

6.9 Switching Characteristics

$V_{CC} = 4.5\text{ V to }15\text{ V}$, $T_A = -40^\circ\text{C to }125^\circ\text{C}$ (unless otherwise noted) (see [Figure 1](#))

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_R Rise time (OUTA, OUTB)	$C_{LOAD} = 1.8\text{ nF}$		20	40	ns
t_F Fall time (OUTA, OUTB)	$C_{LOAD} = 1.8\text{ nF}$		15	40	ns
t_{D1} Delay time, IN rising (IN to OUT)	$C_{LOAD} = 1.8\text{ nF}$		25	40	ns
t_{D2} Delay time, IN falling (IN to OUT)	$C_{LOAD} = 1.8\text{ nF}$		35	50	ns

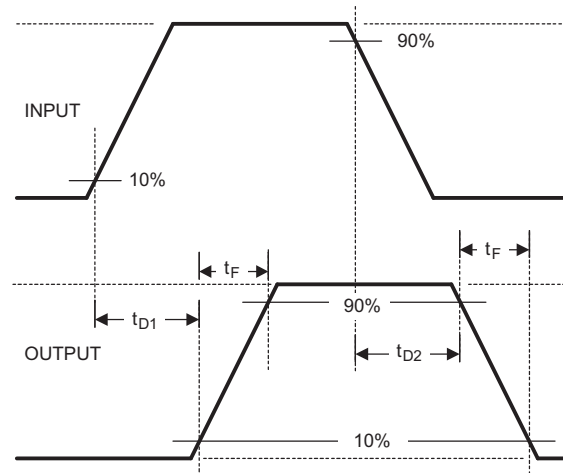


Figure 1. Switching Waveforms for Noninverting Driver

6.10 Typical Characteristics

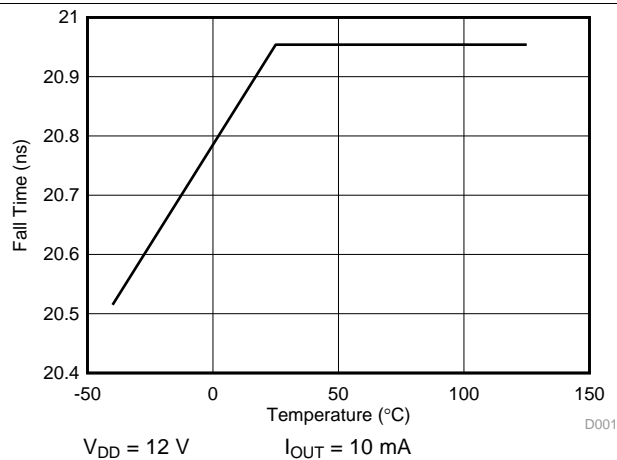


Figure 2. Output Pulldown Resistance vs Temperature

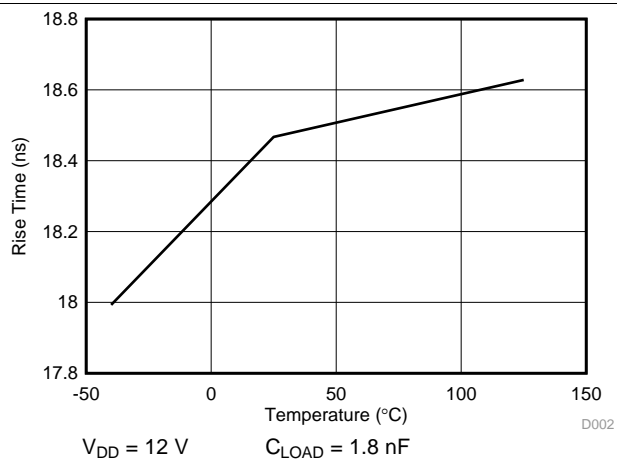


Figure 3. Rise Time vs Temperature

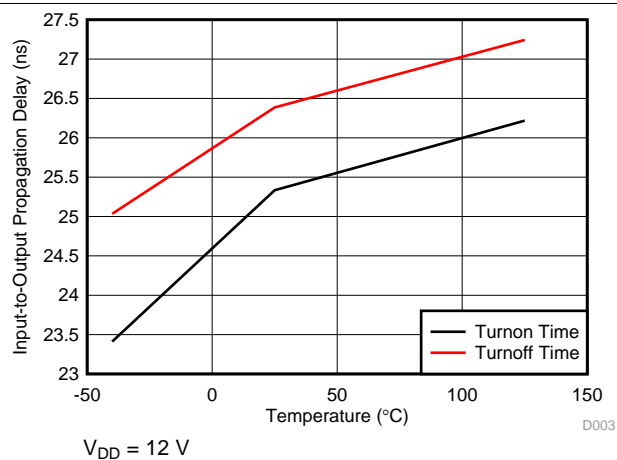


Figure 4. Input to Output Propagation Delay vs Temperature

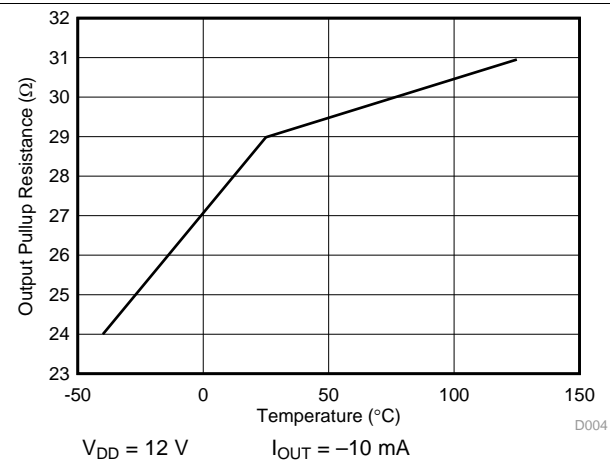


Figure 5. Output Pullup Resistance vs Temperature

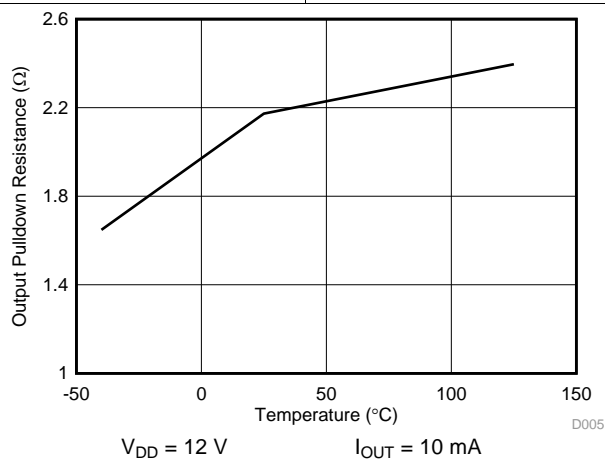


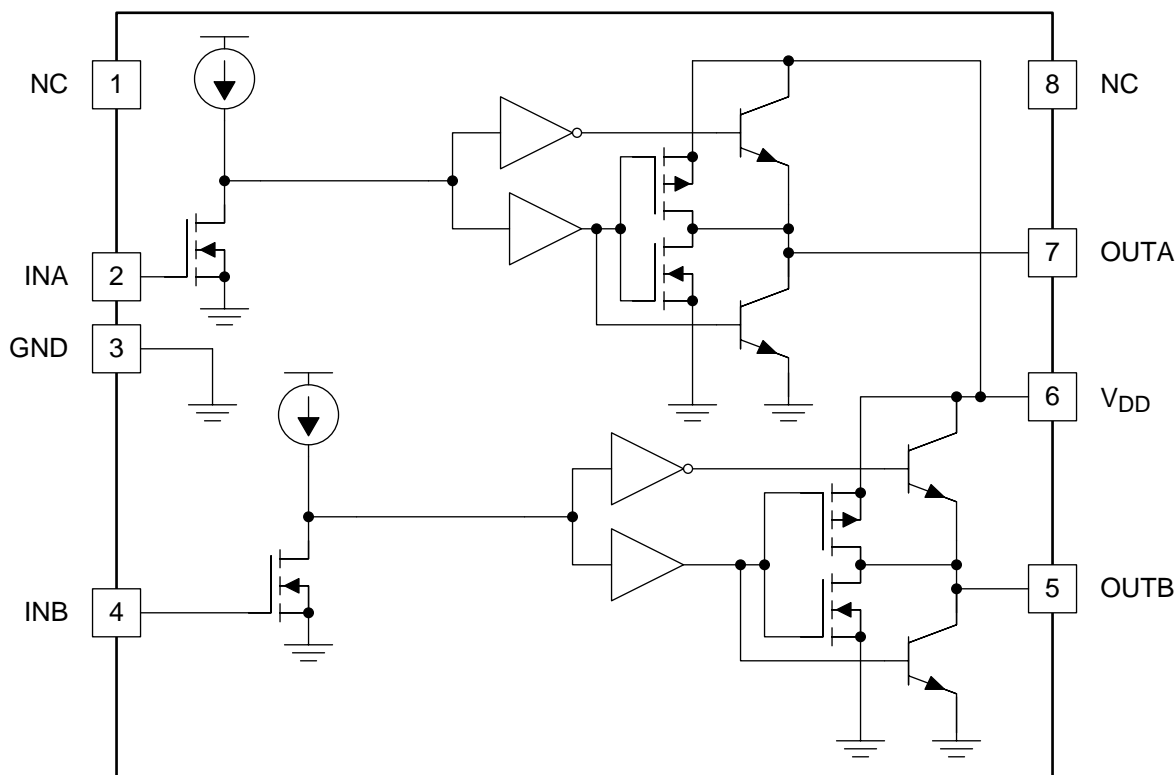
Figure 6. Output Pulldown Resistance vs Temperature

7 Detailed Description

7.1 Overview

The UCC27324-Q1 device represents Texas Instruments' latest generation of dual-channel, low-side, high-speed gate-driver devices featuring a 4-A source and sink capability. With industry leading switching characteristics, automotive qualification, and a host of other features shown on the first page, the UCC27324-Q1 provides an efficient, robust, and reliable solution to your high current low-side driver needs in automotive applications.

7.2 Functional Block Diagram



7.3 Feature Description

7.3.1 Input Stage

The input thresholds have a 3.3-V logic sensitivity over the full range of V_{DD} voltage, yet it is equally compatible with 0 V to V_{DD} signals.

The inputs of UCC27324-Q1 device are designed to withstand 500-mA reverse current without damage to the device or logic upset. The input stage of each driver should be driven by a signal with a short rise or fall time. This condition is satisfied in typical power-supply applications, where the input signals are provided by a PWM controller or logic gates with fast transition times (<200 ns). The input stages to the drivers function as a digital gate, and they are not intended for applications where a slow changing input voltage is used to generate a switching output when the logic threshold of the input section is reached. While this may not be harmful to the driver, the output of the driver may switch repeatedly at a high frequency.

Users should not attempt to shape the input signals to the driver in an attempt to slow down (or delay) the signal at the output. If limiting the rise or fall times to the power device is desired, limit the rise or fall times to the power device, then an external resistance can be added between the output of the driver and the load device, which is generally a power MOSFET gate. The external resistor also may help remove power dissipation from the device package, as discussed in the [Thermal Considerations](#) section.

Feature Description (continued)

7.3.2 Output Stage

Noninverting outputs of the UCC27324-Q1 are intended to drive external N-channel MOSFETs.

Each output stage is capable of supplying ± 4 -A peak current pulses and swings to both V_{DD} and GND. The pullup and pulldown circuits of the driver are constructed of bipolar and MOSFET transistors in parallel. The peak output current rating is the combined current from the bipolar and MOSFET transistors. The output resistance is the $R_{DS(ON)}$ of the MOSFET transistor when the voltage on the driver output is less than the saturation voltage of the bipolar transistor. Each output stage also provides a very low impedance to overshoot and undershoot, due to the body diode of the external MOSFET. This means that, in many cases, external Schottky-clamp diodes are not required.

The UCC27324-Q1 device delivers a 4-A gate drive when it is most needed during the MOSFET switching transition—at the Miller plateau region—providing improved efficiency gains. A unique bipolar and MOSFET hybrid output stage in parallel also allows efficient current sourcing at low supply voltages.

7.4 Device Functional Modes

[Table 1](#) lists the device functions.

Table 1. Function Table

INPUTS		UCC27324-Q1 OUTPUTS	
INA	INB	OUTA	OUTB
L	L	L	L
L	H	L	H
H	L	H	L
H	H	H	H

8 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

High-frequency power supplies often require high-speed high-current drivers such as the UCC27324-Q1 device. A leading application is the need to provide a high-power buffer stage between the PWM output of the control device and the gates of the primary power MOSFET or IGBT switching devices. In other cases, the driver is used to drive the power device gates through a drive transformer. Synchronous rectification supplies must also simultaneously drive multiple devices, which can present an extremely large load to the control circuitry.

Drivers are used when using the primary PWM regulator to directly drive the switching devices is not feasible for one or more reasons. The PWM device may not have the brute drive capability required for the intended switching MOSFET, limiting the switching performance in the application. In other cases, minimizing the effect of high-frequency switching noise by placing the high-current driver physically close to the load may be necessary. Also, newer devices that target the highest operating frequencies may not incorporate onboard gate drivers at all. The PWM outputs are intended to drive only the high-impedance input to a driver such as the UCC27324-Q1. Finally, the control device may be under thermal stress because power dissipation, and an external driver can help by moving the heat from the controller to an external package.

8.1.1 Parallel Outputs

The A and B drivers may be combined into a single driver by connecting the INA and INB inputs together and the OUTA and OUTB outputs together (respectively). Then, a single signal can control the paralleled combination as shown in [Figure 7](#).

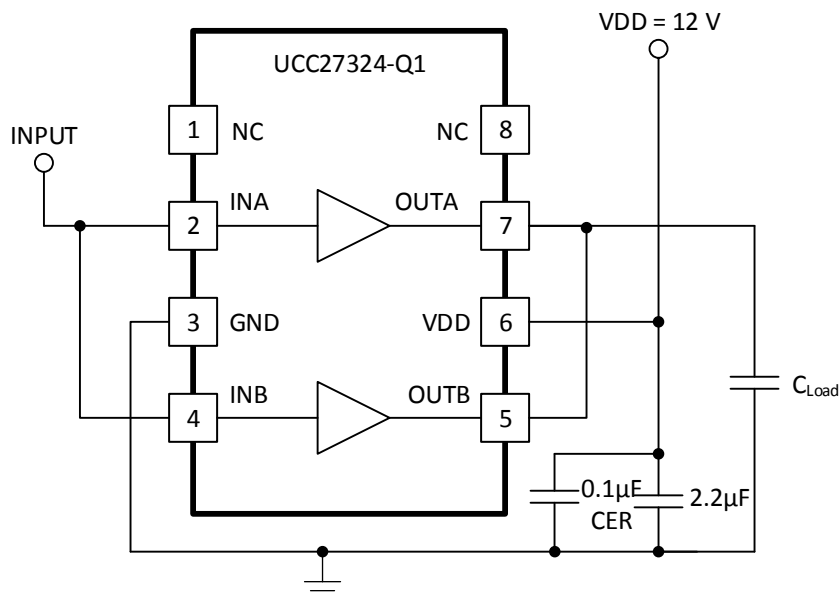


Figure 7. Combined Input and Output Configuration

8.2 Typical Application

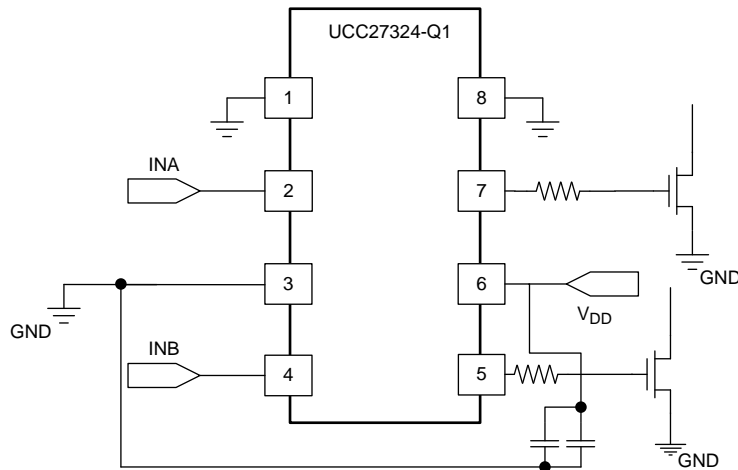


Figure 8. Typical Application Schematic

8.2.1 Design Requirements

When selecting the proper gate driver device for an end application, some design considerations must be evaluated first to make the most appropriate selection. These design considerations include V_{DD} , drive current, and power dissipation.

8.2.2 Detailed Design Procedure

8.2.2.1 Propagation Delay

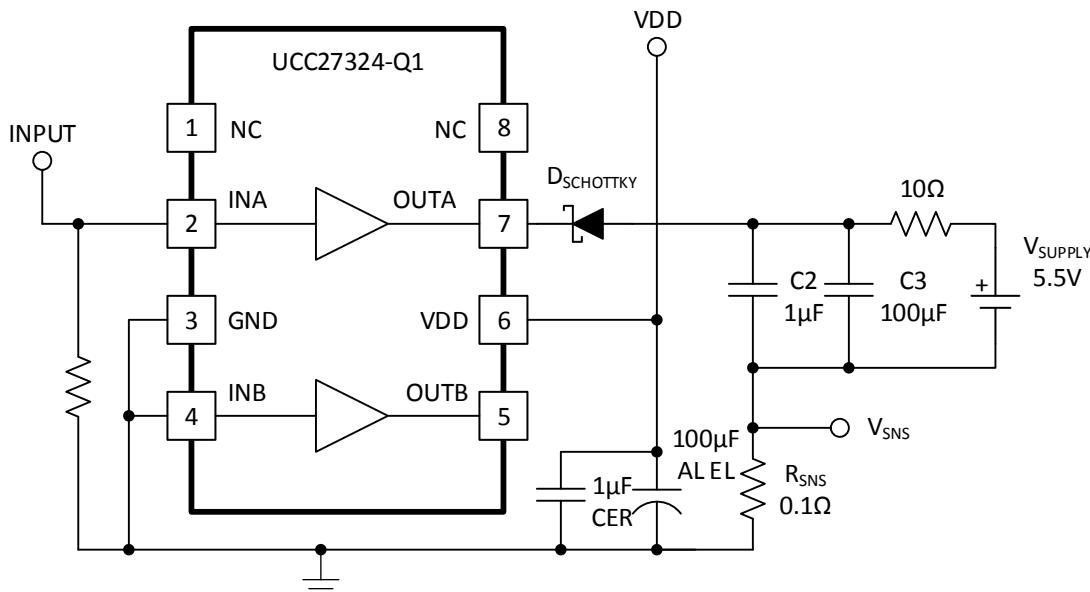
The acceptable propagation delay from the gate driver is dependent on the switching frequency at which it is used and the acceptable level of pulse distortion to the system. The UCC27324-Q1 device features fast 25-ns propagation delays which ensures very little pulse distortion and allows operation at high frequencies. See the [Switching Characteristics](#) table for the propagation and switching characteristics of the UCC27324-Q1 device.

8.2.2.2 Source and Sink Capabilities During Miller Plateau

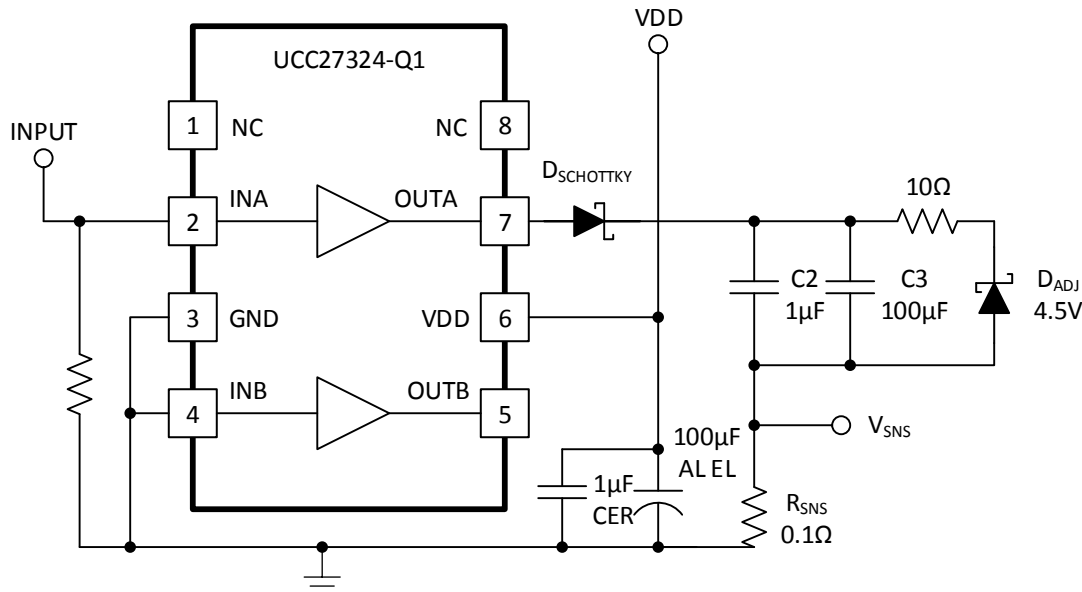
Large power MOSFETs present a large load to the control circuitry. Proper drive is required for efficient, reliable operation. The UCC27324-Q1 drivers have been optimized to provide maximum drive to a power MOSFET during the Miller plateau region of the switching transition. This interval occurs while the drain voltage is swinging between the voltage levels dictated by the power topology, requiring the charging or discharging of the drain-gate capacitance with current supplied or removed by the driver.

Two circuits are used to test the current capabilities of the UCC27324-Q1 driver. In each case, external circuitry is added to clamp the output near 5 V while the device is sinking or sourcing current. An input pulse of 250 ns is applied at a frequency of 1 kHz in the proper polarity for the respective test. In each test, there is a transient period when the current peaked up and then settled down to a steady-state value. The noted current measurements are made at a time of 200 ns after the input pulse is applied, after the initial transient.

The circuit in [Figure 9](#) is used to verify the current sink capability when the output of the driver is clamped at approximately 5 V, a typical value of gate-source voltage during the Miller plateau region. The UCC27324-Q1 is found to sink 4.5 A at $V_{DD} = 15$ V and 4.28 A at $V_{DD} = 12$ V.

Typical Application (continued)

Figure 9. Current Sink Test Circuit

The circuit in [Figure 10](#) is used to test the current source capability with the output clamped to approximately 5 V with a string of Zener diodes. The UCC27324-Q1 device can source 4.8 A at $V_{DD} = 15$ V and 3.7 A at $V_{DD} = 12$ V.


Figure 10. Current Source Test Circuit
NOTE

The current-sink capability is slightly stronger than the current source capability at lower V_{DD} because of the differences in the structure of the bipolar-MOSFET power output section, where the current source is a P-channel MOSFET and the current sink has an N-channel MOSFET.

Typical Application (continued)

In a large majority of applications, the fact that the turn-off capability of a driver is stronger than the turn-on capability is advantageous which helps to ensure that the MOSFET is held off during common power-supply transients that may turn the device back on.

8.2.2.3 Supply Voltage (V_{DD})

Although quiescent V_{DD} current is very low, total supply current is higher, depending on the OUTA and OUTB current and the programmed oscillator frequency. The total V_{DD} current is the sum of quiescent V_{DD} current and the average OUT current. With the known operating frequency and the MOSFET gate charge (Q_g), use [Equation 1](#) to calculate the average OUT current.

$$I_{OUT} = Q_g \times f$$

where

- f is frequency (1)

For the best high-speed circuit performance, two V_{DD} bypass capacitors are recommended to prevent noise problems. The use of surface mount components is highly recommended. A 0.1- μ F ceramic capacitor must be located closest to the V_{DD} to ground connection. In addition, a larger capacitor (such as 1- μ F) with relatively low ESR must be connected in parallel, to help deliver the high current peaks to the load. The parallel combination of capacitors must present a low impedance characteristic for the expected current levels in the driver application.

8.2.2.4 Drive Current and Power Requirements

The UCC27324-Q1 drivers are capable of delivering 4 A of current to a MOSFET gate for a period of several hundred nanoseconds. High peak current is required to quickly turn on the device. Then, to turn off the device, the driver is required to sink a similar amount of current to ground. This repeats at the operating frequency of the power device. A MOSFET is used in this discussion, because it is the most common type of switching device used in high-frequency power-conversion equipment.

Reference [\[1\]](#) in the [相关文档](#) section discuss the current required to drive a power MOSFET and other capacitive-input switching devices and includes information on the previous generation of bipolar gate drivers.

When a driver is tested with a discrete capacitive load, calculating the power that is required from the bias supply is fairly simple. Use [Equation 2](#) to calculate the energy that must be transferred from the bias supply to charge the capacitor.

$$E = \frac{1}{2}CV^2$$

where

- C is the load capacitor
- V is the bias voltage feeding the driver (2)

An equal amount of energy transferred to ground when the capacitor is discharged which leads to power loss. Use [Equation 3](#) to calculate this power loss.

$$P = 2 \times \frac{1}{2}CV^2f$$

where

- f is the switching frequency (3)

This power is dissipated in the resistive elements of the circuit. Thus, with no external resistor between the driver and gate, this power is dissipated inside the driver. Half of the total power is dissipated when the capacitor is charged, and the other half is dissipated when the capacitor is discharged. An actual example using the conditions of the previous gate drive waveform should help clarify this.

Use [Equation 4](#) to calculate the power loss with the following values: $V_{DD} = 12$ V, $C_{LOAD} = 10$ nF, and $f = 300$ kHz.

$$P = 10 \text{ nF} \times (12)^2 \times (300 \text{ kHz}) = 0.432 \text{ W} \quad (4)$$

For a 12-V supply, use [Equation 5](#) to calculate the current

$$I = P / V = 0.432 \text{ W} / 12 \text{ V} = 0.036 \text{ A} \quad (5)$$

Typical Application (continued)

The actual current measured from the supply was 0.037 A, which is very close to the predicted value. But, the I_{DD} current that is due to the internal consumption should be considered. With no load, the current draw is 0.0027 A. Under this condition, the output rise and fall times are faster than with a load. This could lead to an almost insignificant, yet measurable current due to cross-conduction in the output stages of the driver. However, these small current differences are buried in the high-frequency switching spikes and are beyond the measurement capabilities of a basic lab setup. The measured current with 10-nF load is reasonably close to the expected value.

The switching load presented by a power MOSFET can be converted to an equivalent capacitance by examining the gate charge required to switch the device. This gate charge includes the effects of the input capacitance plus the added charge needed to swing the drain of the device between the on and off states. Most manufacturers provide specifications that provide the typical and maximum gate charge, in nC, to switch the device under specified conditions. Using the gate charge Q_g , one can determine the power that must be dissipated when charging a capacitor. Use Equation 6 and the equivalence $Q_g = C_{eff}V$ to calculate this power.

$$P = C \times V^2 \times f = V \times Q_g \times f \quad (6)$$

Equation 6 allows a power designer to calculate the bias power required to drive a specific MOSFET gate at a specific bias voltage.

8.2.3 Application Curve

Figure 11 shows the circuit performance achievable with a single driver (half of the 8-pin device) driving a 10-nF load. The input pulse width (not shown) is set to 300 ns to show both transitions in the output waveform. The rising and falling edges of the switching waveforms are fairly linear which is because of the constant output current characteristic of the driver as opposed to the resistive output impedance of traditional MOSFET-based gate drivers.

Sink and source currents of the driver are dependent upon the V_{DD} value and the output capacitive load. The larger the V_{DD} value the higher the current capability, and the larger the capacitive load the higher the current sink/source capability. Trace resistance and inductance, including wires and cables for testing, slows down the rise and fall times of the outputs which reduces the current capabilities of the driver. To achieve higher current results, reduce resistance and inductance on the board as much as possible and increase the capacitive output load value to swap out the effect of the inductance values.

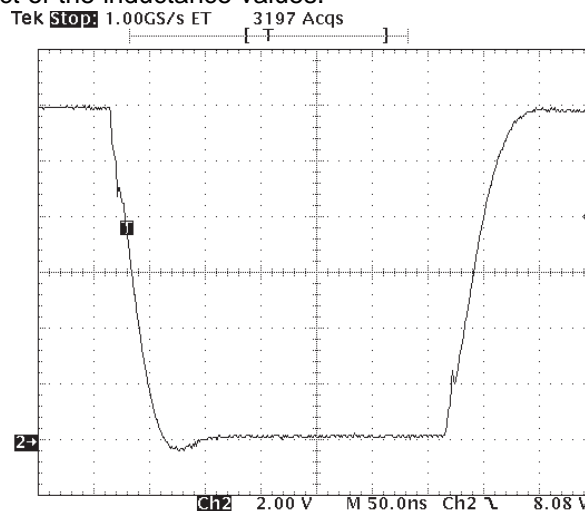


Figure 11. Single Driver With 10-nF Load, 300-ns Pulse-Width Input

9 Power Supply Recommendations

The bias supply voltage range for which the UCC27324-Q1 device is rated to operate is from 4 V to 15 V. A quiet, robust power supply capable of delivering at least 4 A should be used to ensure proper operation of the device.

10 Layout

10.1 Layout Guidelines

For the best high-speed circuit performance, two V_{DD} bypass capacitors are recommended to prevent noise problems. The use of surface-mount components is highly recommended. A 0.1- μ F ceramic capacitor should be located closest to the V_{DD} to ground connection. In addition, a larger capacitor (such as 1- μ F) with relatively low ESR should be connected in parallel, to help deliver the high current peaks to the load. The parallel combination of capacitors should present a low-impedance characteristic for the expected current levels in the driver application.

In a power driver operating at high frequency, it is a significant challenge to get clean waveforms without much overshoot/undershoot and ringing. The low output impedance of these drivers produces waveforms with high $\Delta i/\Delta t$. This tends to induce ringing in the parasitic inductances. Utmost care must be used in the circuit layout. It is advantageous to connect the driver as close as possible to the leads. The driver layout has ground on the opposite side of the output, so the ground should be connected to the bypass capacitors and the load with copper trace as wide as possible. These connections also should be made with a small enclosed loop area to minimize the inductance.

PCB layout is a critical step in the production process in high-current fast-switching circuits to ensure appropriate operation and design robustness. The UCC27324-Q1 MOSFET driver is capable of delivering large current peaks with rapid rise and fall times at the gate of a power MOSFET to facilitate voltage transitions quickly. At higher V_{DD} voltages, the peak current capability is even higher. High di/dt causes unacceptable ringing if the trace lengths and impedances are not well controlled.

- Locate the driver device as close as possible to the power device in order to minimize the length of high-current traces between the output pins and the gate of the MOSFET being driven.
- Locate the V_{DD} bypass capacitors between V_{DD} and GND as close as possible to the driver with minimal trace length to improve the noise filtering. Place these capacitors as close to each other as is allowed, as shown by C1 and C2 in [Figure 12](#) which ensures minimal trace inductance and gives the effect of a *capacitor bank*. These capacitors support high peak current being drawn from V_{DD} during turn-on of the power MOSFET. The use of low inductance surface mount components is highly recommended.
- Star-point grounding is a good way to minimize noise coupling from one current loop to another. The GND of the driver is connected the other circuit nodes such as the source of the power MOSFET and ground of the PWM controller at one single point. The connected paths must be as short as possible to reduce inductance and be as wide as possible to reduce resistance.
- Use a ground plane to provide noise shielding. Fast rise and fall times at OUT may corrupt the input signals during transition. The ground plane must not be a conduction path for any current loop. Instead, the ground plane must be connected to the star-point with one single trace to establish the ground potential. In addition to noise shielding, the ground plane can help with thermal protection.
- Tie pins 1 and 8 to GND to eliminate any chance of noise causing malfunction on a floating node.

10.2 Layout Example

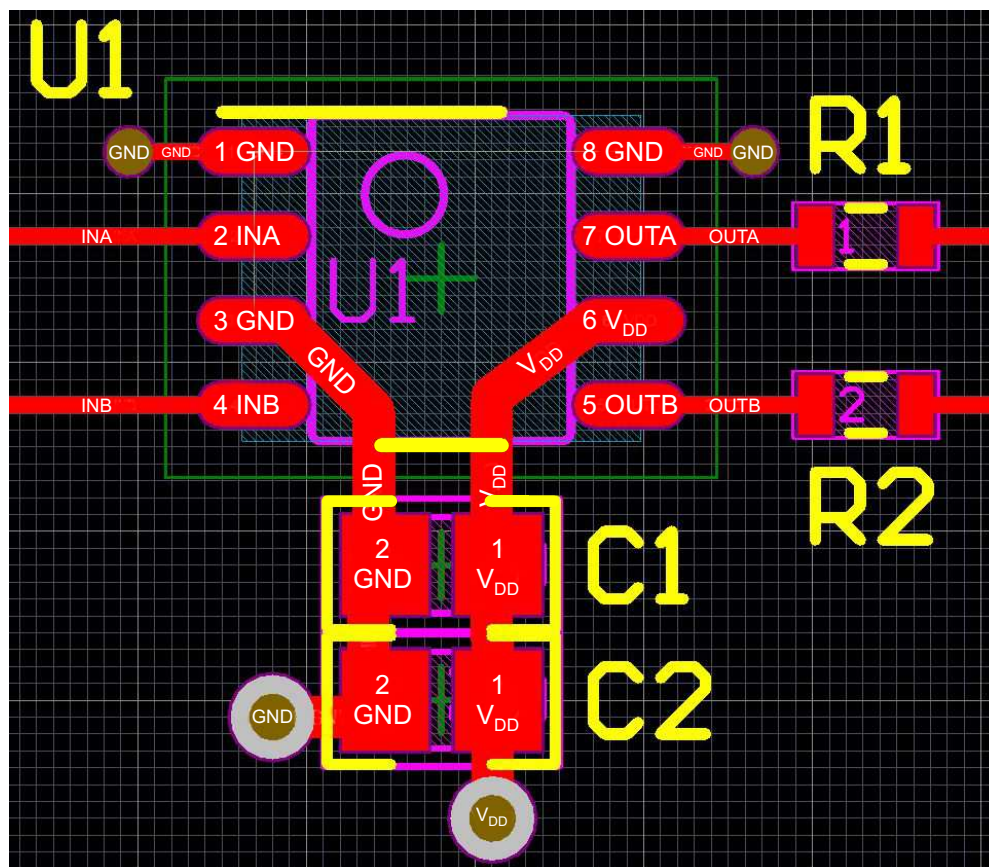


Figure 12. UCC27324-Q1 Layout Example

10.3 Thermal Considerations

The useful range of a driver is greatly affected by the drive power requirements of the load and the thermal characteristics of the package. For a power driver to be useful over a particular temperature range, the package must allow for the efficient removal of the heat produced while keeping the junction temperature within rated limits. The UCC27324-Q1 device of drivers is available in three different packages to cover a range of application requirements.

As shown in [Thermal Information](#), the SOIC-8 (D) package has power ratings of approximately 0.5 W at $T_A = 70^\circ\text{C}$. This limit is imposed in conjunction with the power derating factor also given in the table. Note that the power dissipation in our earlier example is 0.432 W with a 10-nF load, 12-V V_{DD} , switched at 300 kHz. Thus, only one load of this size could be driven using the D or P package, even if the two onboard drivers are paralleled. The difficulties with heat removal limit the drive available in the older packages.

11 器件和文档支持

11.1 文档支持

11.1.1 相关文档

请参阅如下相关文档：

1. 高性能 MOSFET、IGBT 和 MCT 栅极驱动电路的实用注意事项, Bill Andreyckak, [SLUA105](#)
2. 《PowerPAD 耐热增强型封装》, [SLMA002](#)
3. 《PowerPAD 速成》, [SLMA004](#)

11.2 社区资源

下列链接提供到 TI 社区资源的连接。链接的内容由各个分销商“按照原样”提供。这些内容并不构成 TI 技术规范，并且不一定反映 TI 的观点；请参阅 TI 的《使用条款》。

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11.4 静电放电警告



这些装置包含有限的内置 ESD 保护。存储或装卸时，应将导线一起截短或将装置放置于导电泡棉中，以防止 MOS 门极遭受静电损伤。

11.5 术语表

[SLYZ022](#) — TI 术语表。

这份术语表列出并解释术语、缩写和定义。

12 机械、封装和可订购信息

以下页面包含机械、封装和可订购信息。这些信息是指定器件的最新可用数据。数据如有变更，恕不另行通知，且不会对此文档进行修订。如需获取此数据表的浏览器版本，请查阅左侧的导航栏。

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
UCC27324QDRQ1	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	27324Q	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSELETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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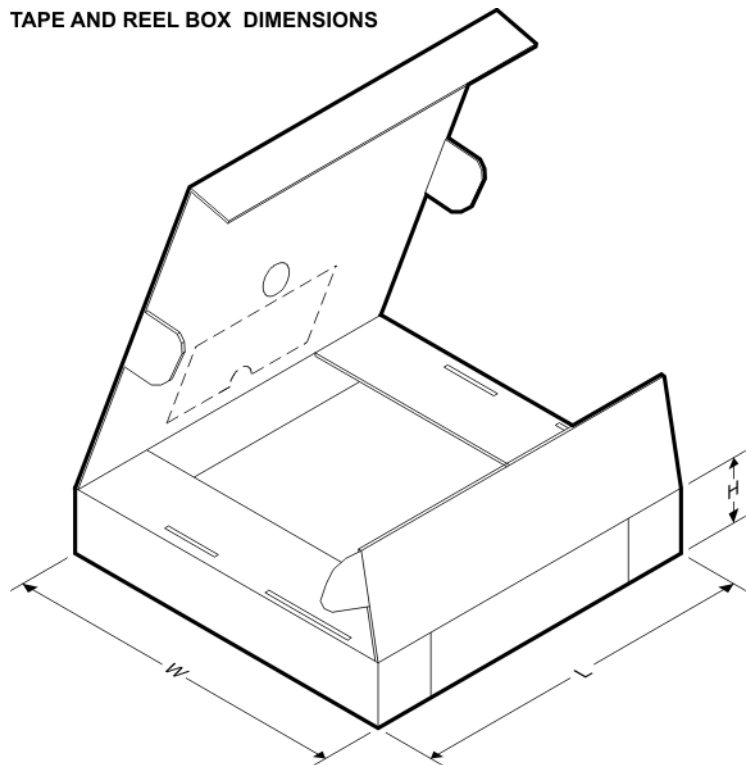
TAPE AND REEL INFORMATION



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
UCC27324QDRQ1	SOIC	D	8	2500	330.0	12.5	6.4	5.2	2.1	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
UCC27324QDRQ1	SOIC	D	8	2500	340.5	336.1	25.0



D0008A

PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



4214825/C 02/2019

NOTES:

1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
4. This dimension does not include interlead flash.
5. Reference JEDEC registration MS-012, variation AA.

EXAMPLE BOARD LAYOUT

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE:8X



SOLDER MASK DETAILS

4214825/C 02/2019

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



SOLDER PASTE EXAMPLE
BASED ON .005 INCH [0.125 MM] THICK STENCIL
SCALE:8X

4214825/C 02/2019

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

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